

Fig. 1A

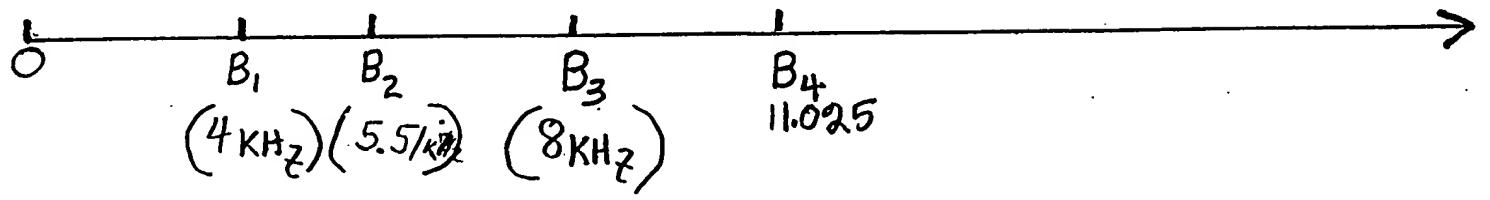


Fig. 1B

FFT-based Scalable and Embedded Codec Architecture

Encoder with M Octave Bands

SCALABLE AND EMBEDDED
CODEC ARCHITECTURE

$$\begin{aligned}
 & \text{2}^M \text{ Hz} \text{ sampled input} \xrightarrow{\text{Windowing \&} (2^{M-1}N)-\text{point FFT}} S_m(n) \\
 & S_m(k) = \text{FFT of windowed } S_m(n) \\
 & \delta_{m1} = \{S_m(k)\}_{k=0}^{\frac{N}{2}-1} \\
 & \delta_{m2} = \{S_m(k)\}_{k=\frac{N}{2}}^{N-1} \\
 & \delta_{m3} = \{S_m(k)\}_{k=N}^{2N-1} \\
 & \vdots \\
 & \delta_{mM} = \{S_m(k)\}_{k=2^{M-3}N}^{2^{M-2}N-1}
 \end{aligned}$$

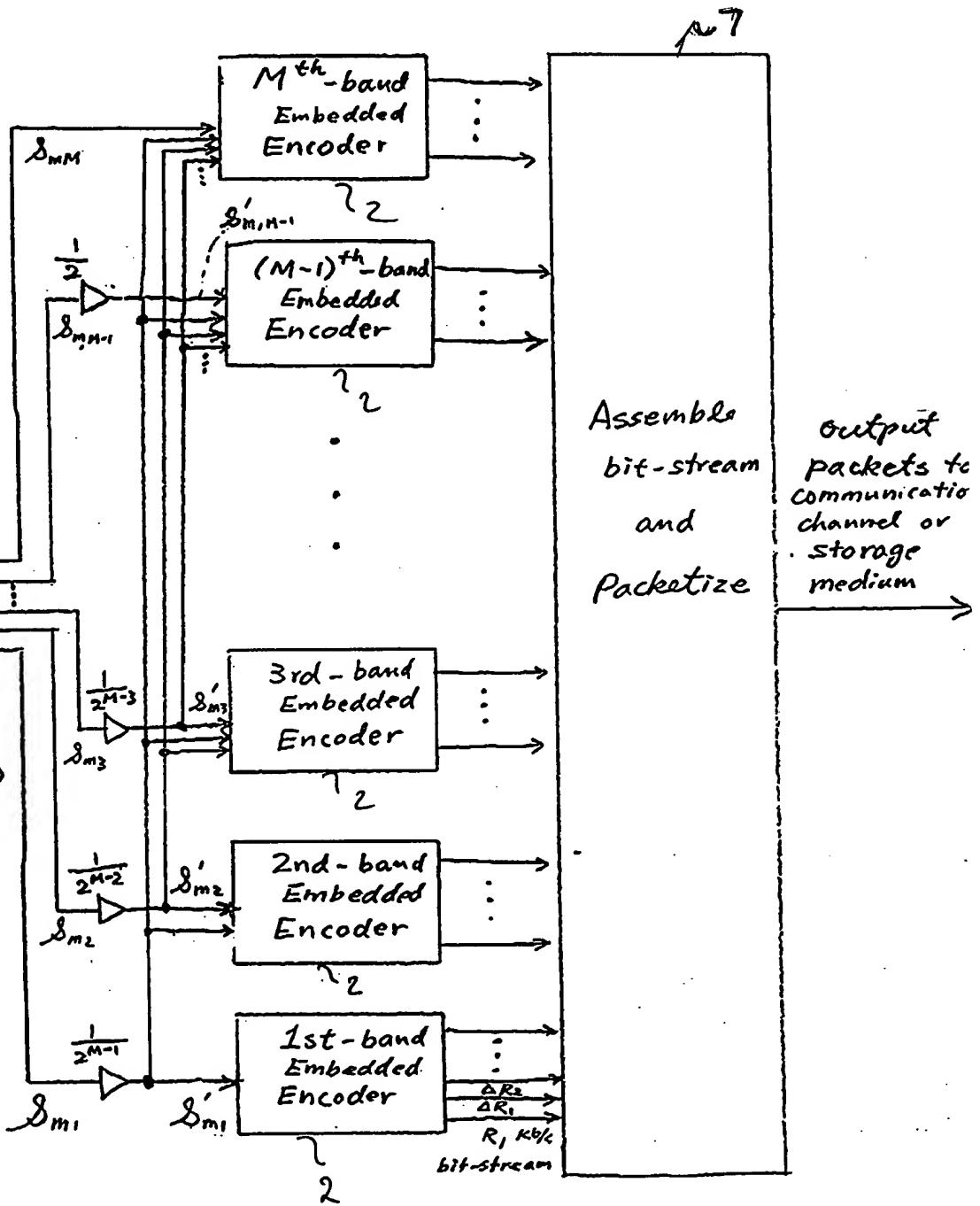
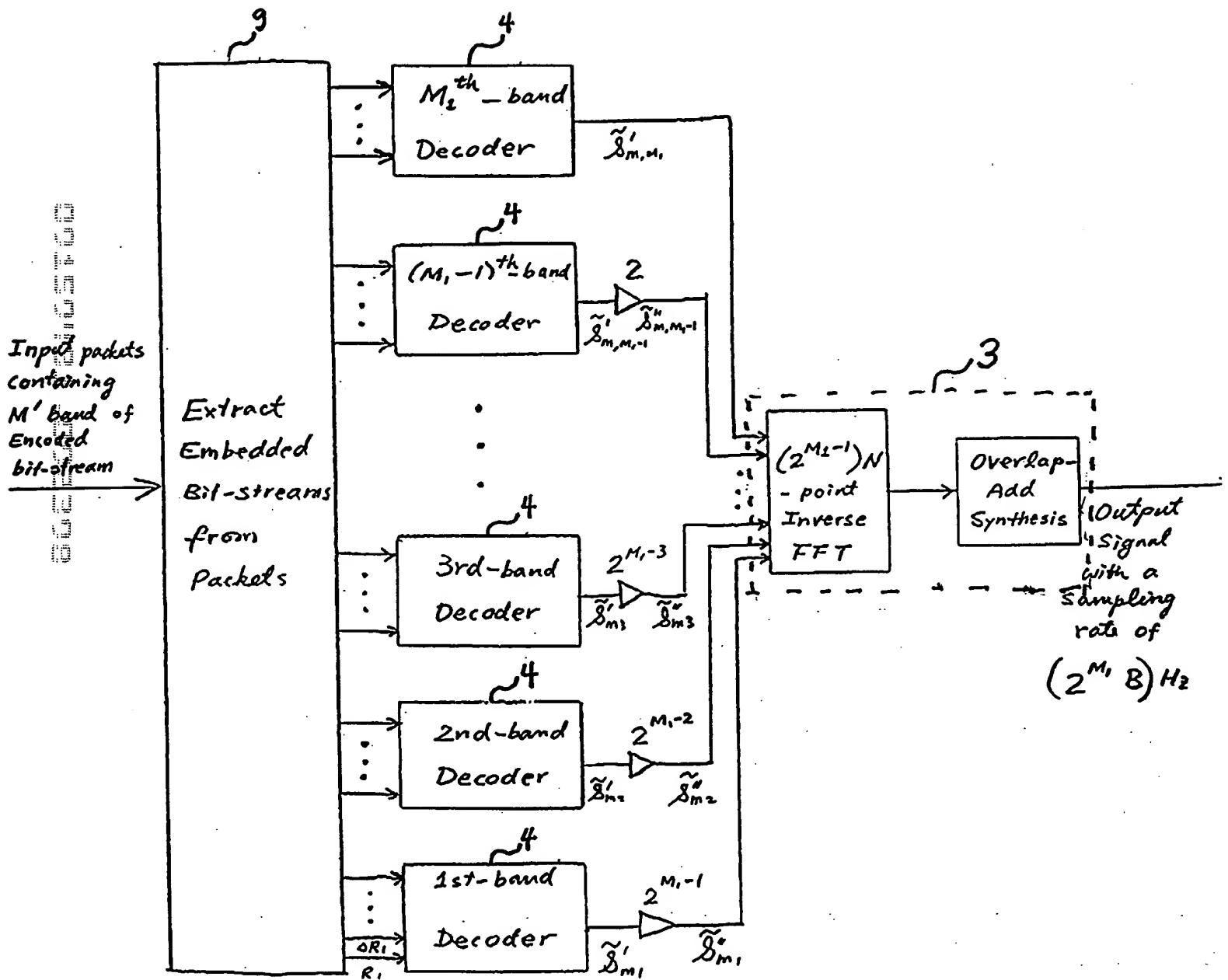


Fig. 2A

FFT-based Scalable and Embedded Codec Architecture

— Decoder with M_1 Octave Bands ($1 \leq M_1 \leq M$)



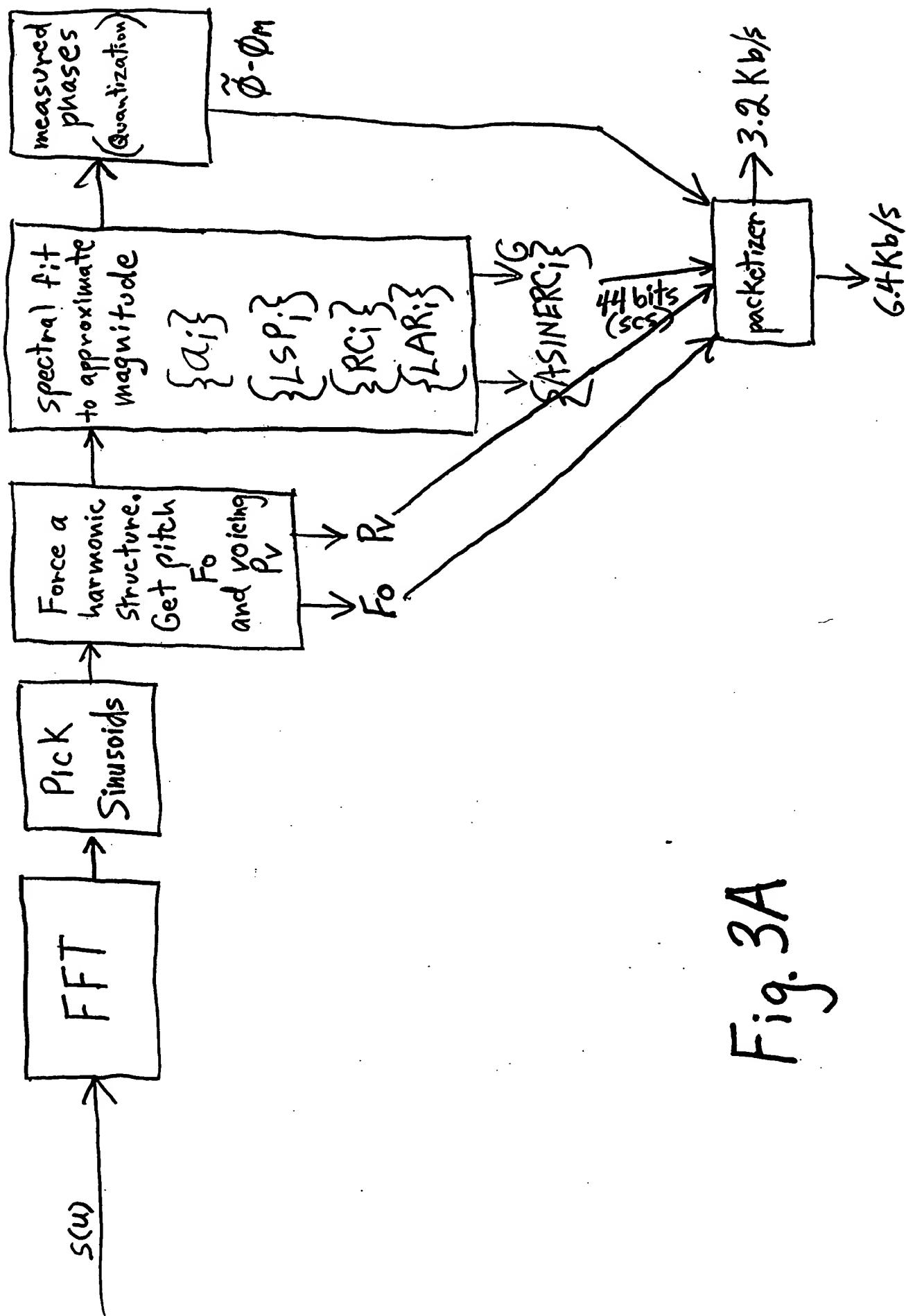
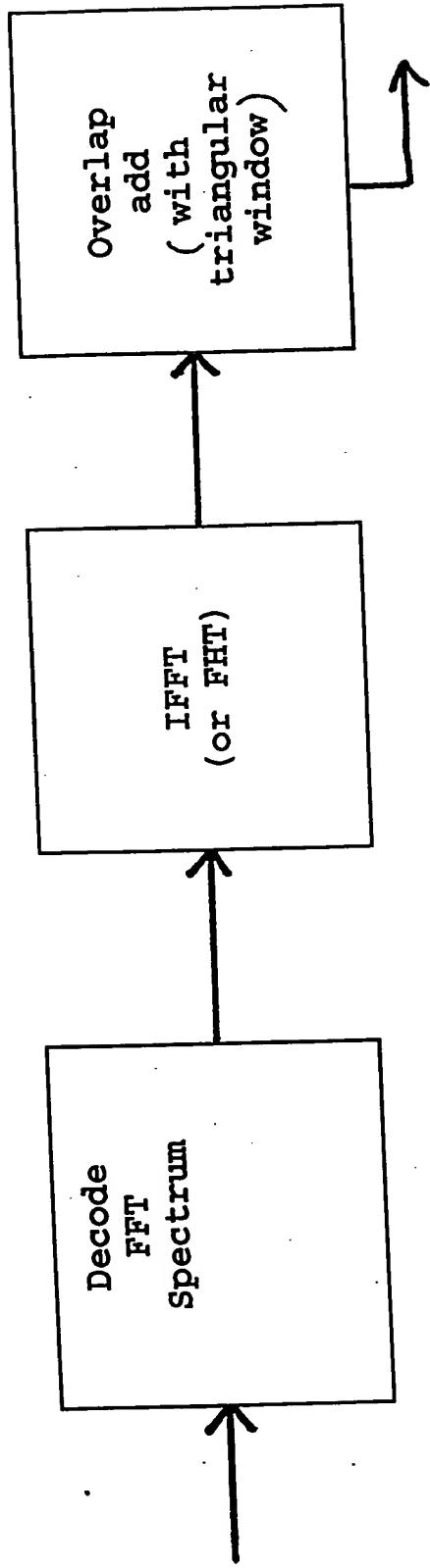


Fig. 3A



Decoder: Synthesis every M ms.

Fig. 3B

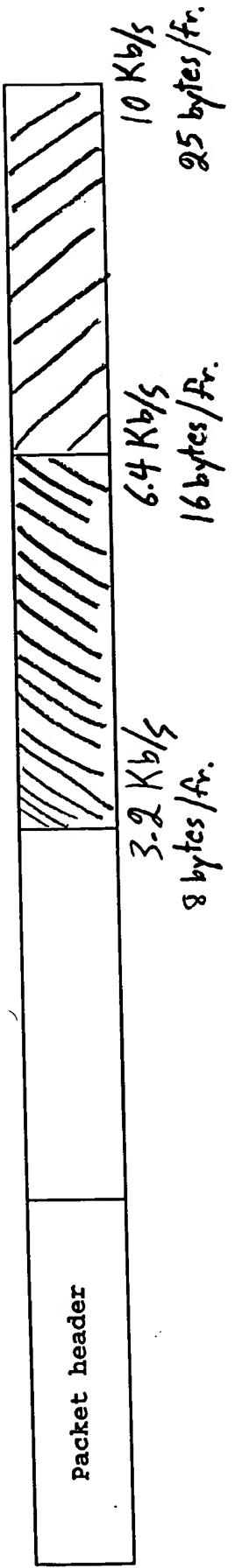
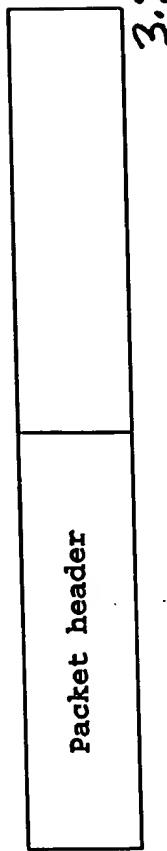
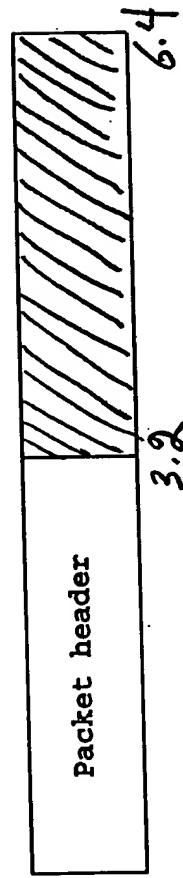


Fig. 4A

1st priority packet



2nd priority packet



3rd priority packet

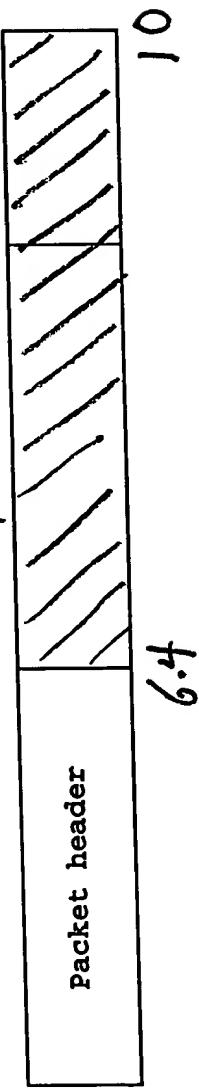


Fig. 4B

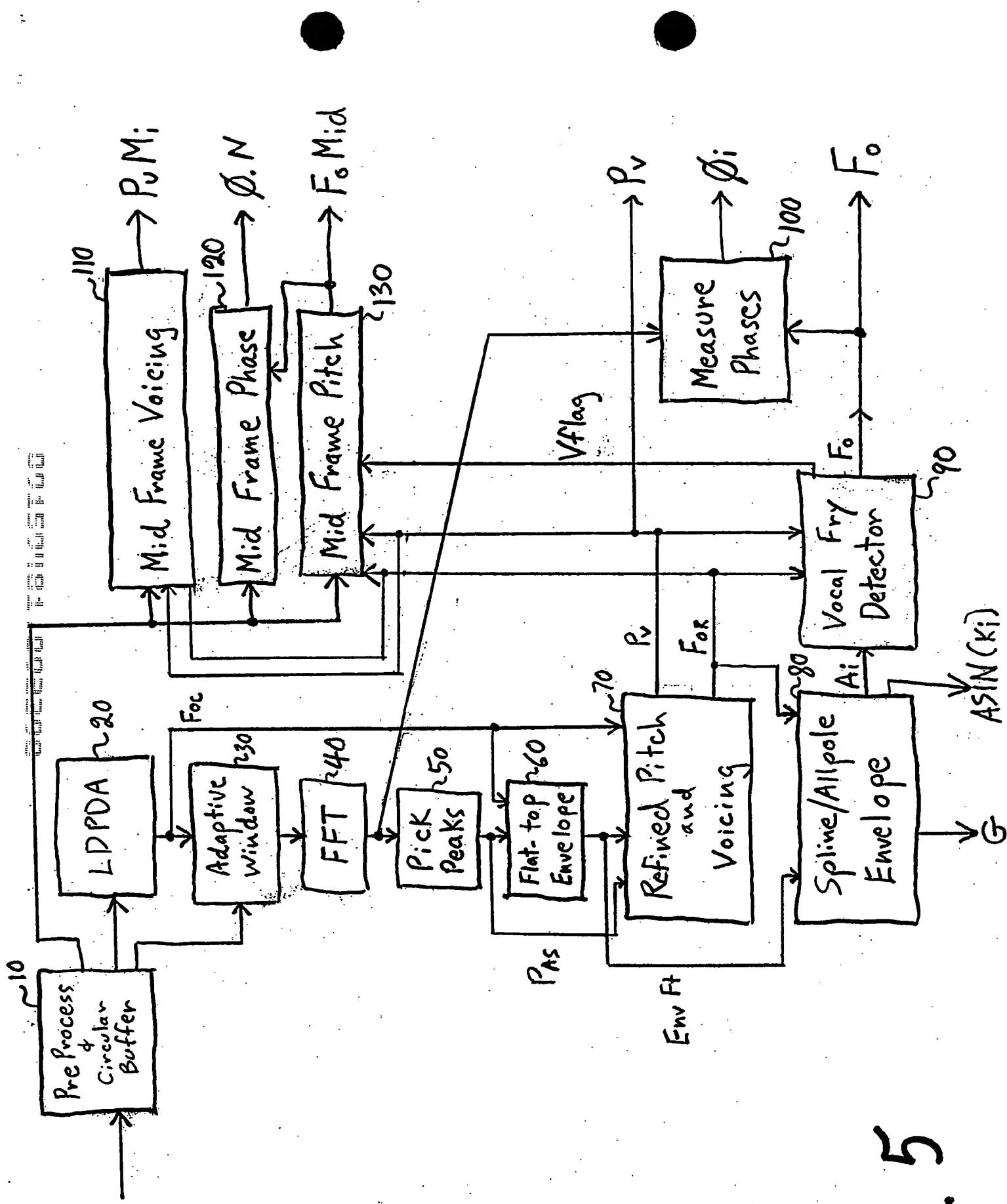


Fig. 5

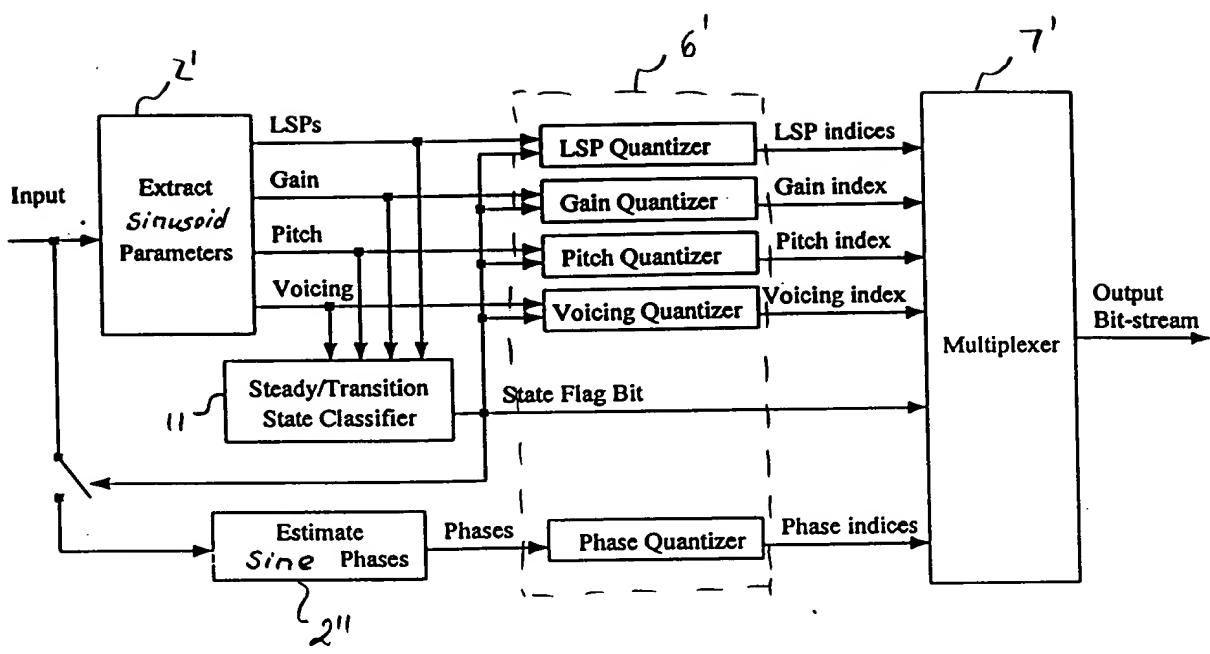


Fig 5A

SPEECH SYNTHESIZER

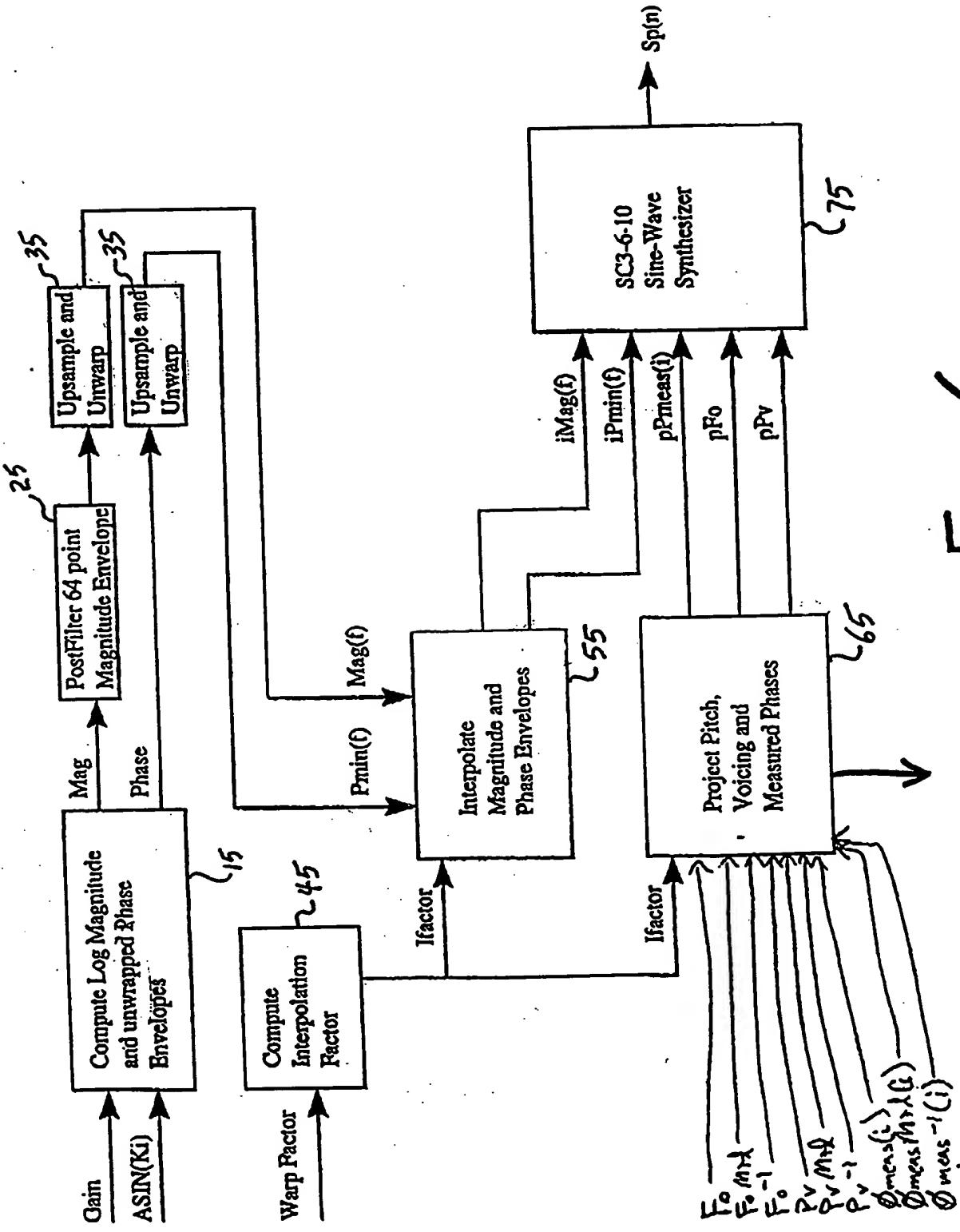


Fig. 6

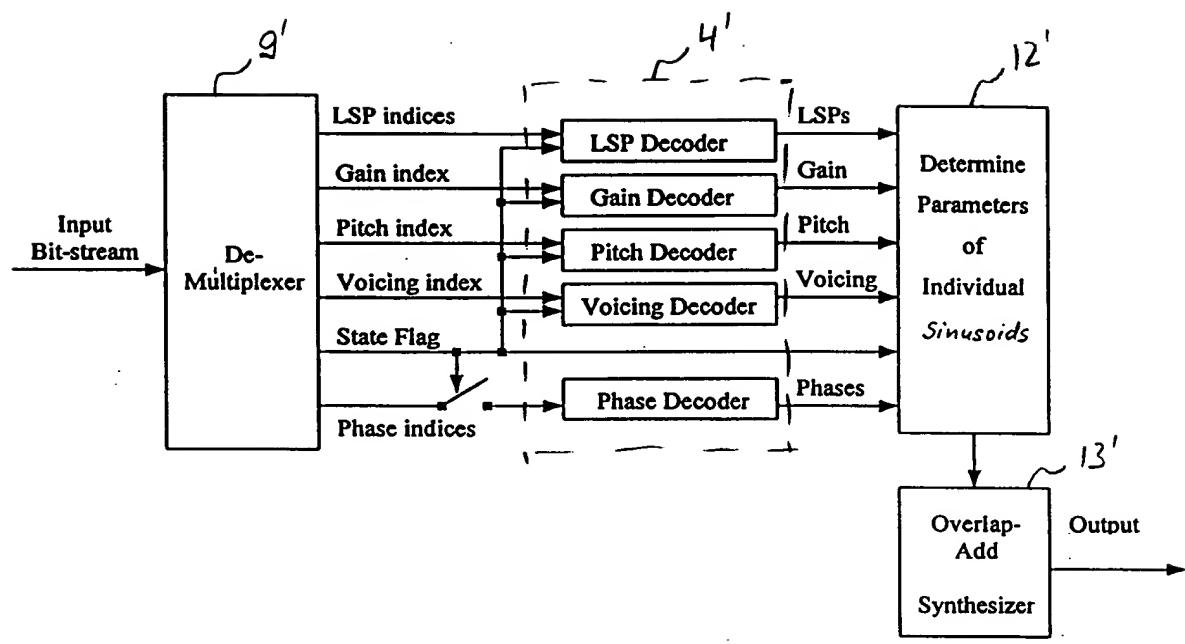


Fig. 6 A

SC3-6-10 SINE WAVE SYNTHESIZER

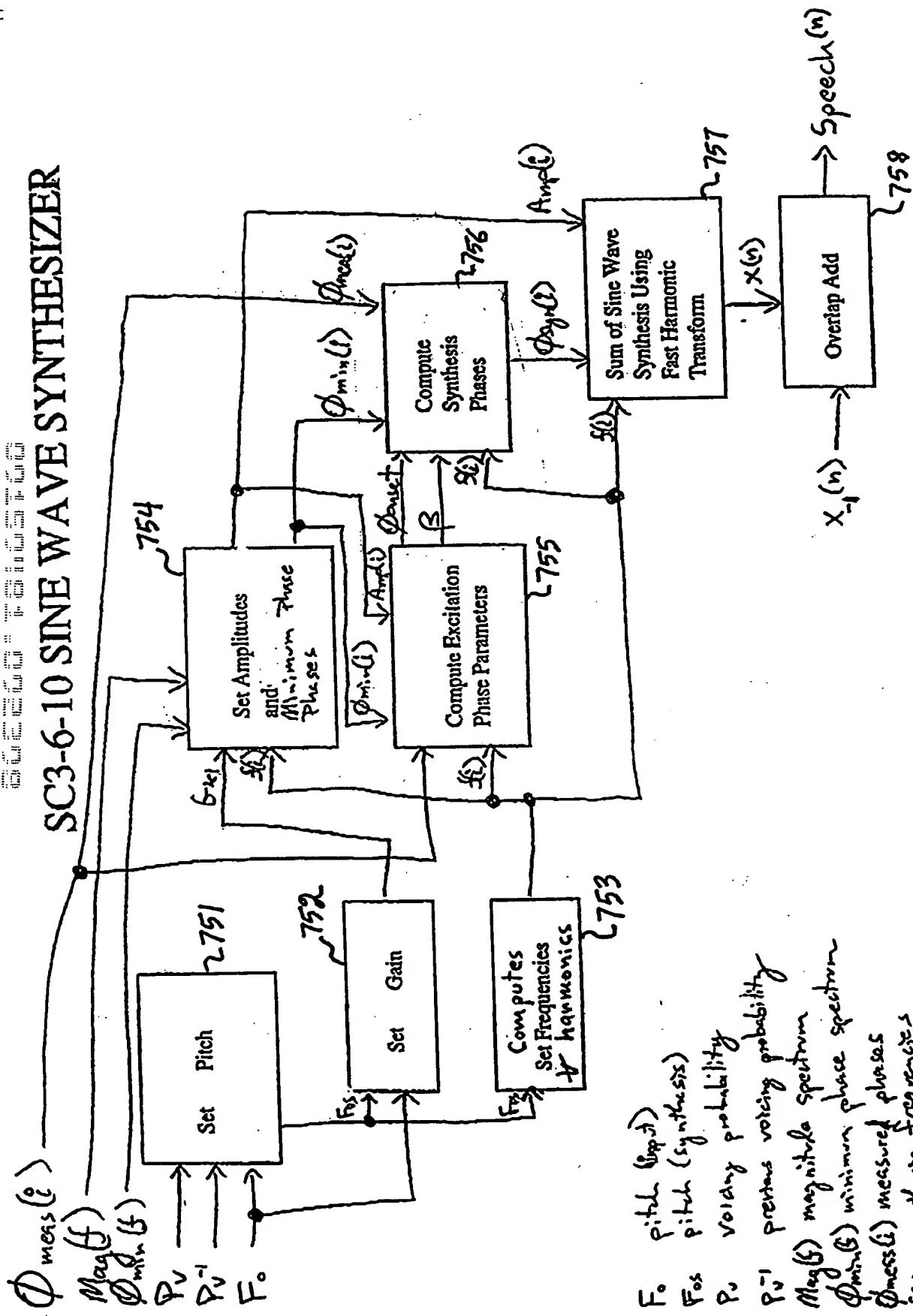
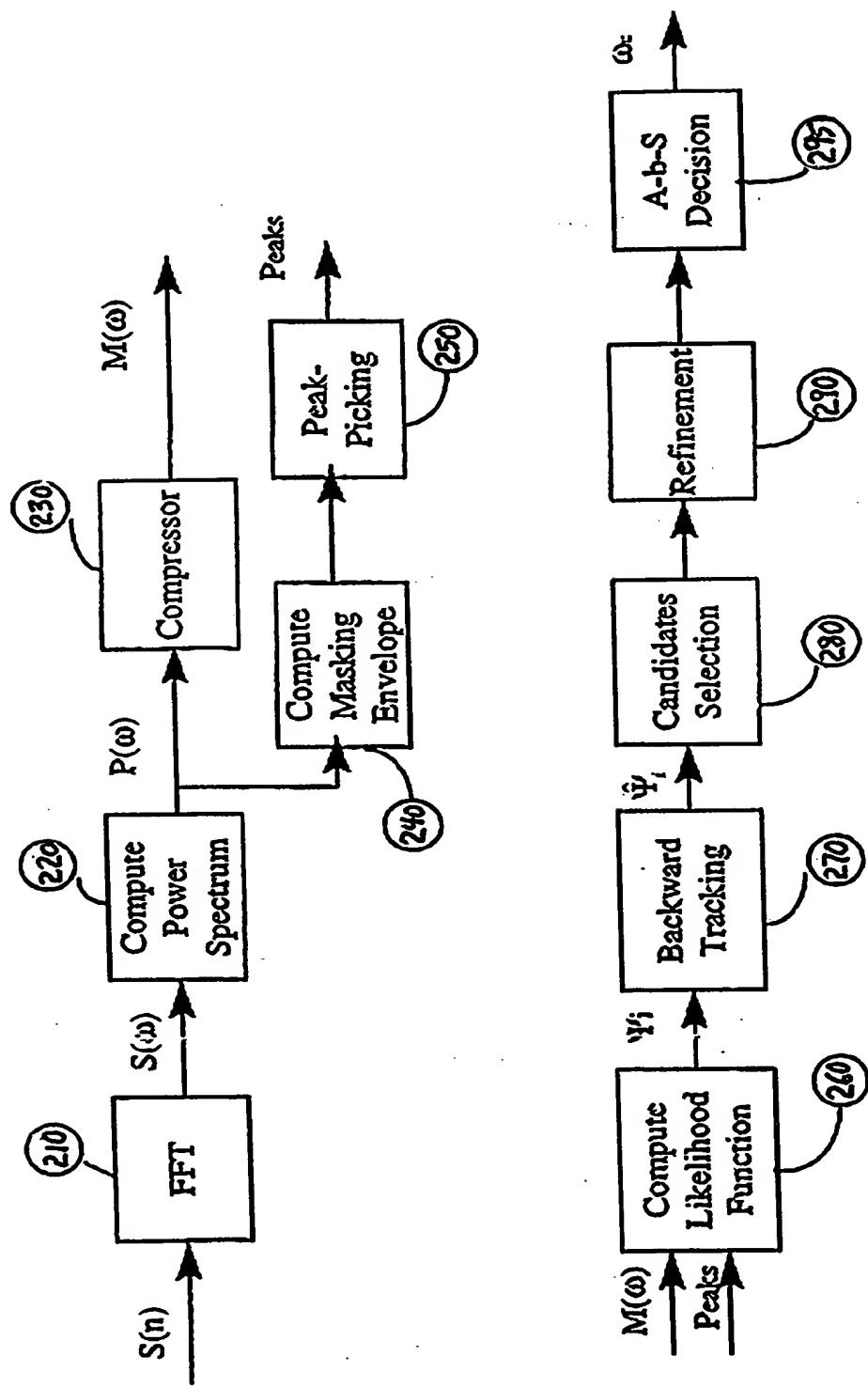


Fig. 7

Fig. 8



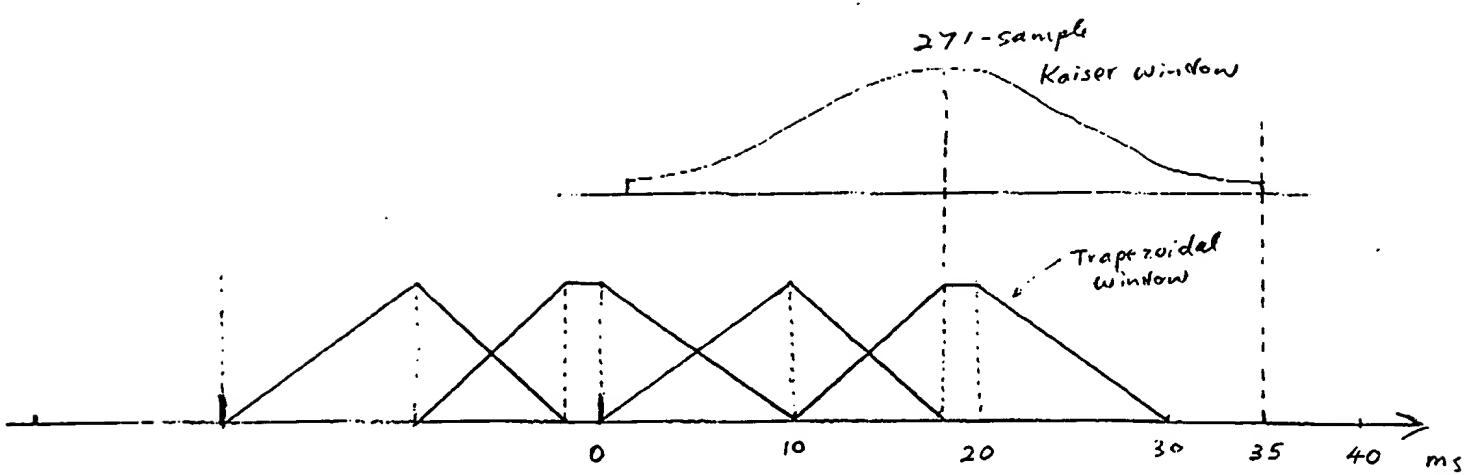


Fig. 8. A

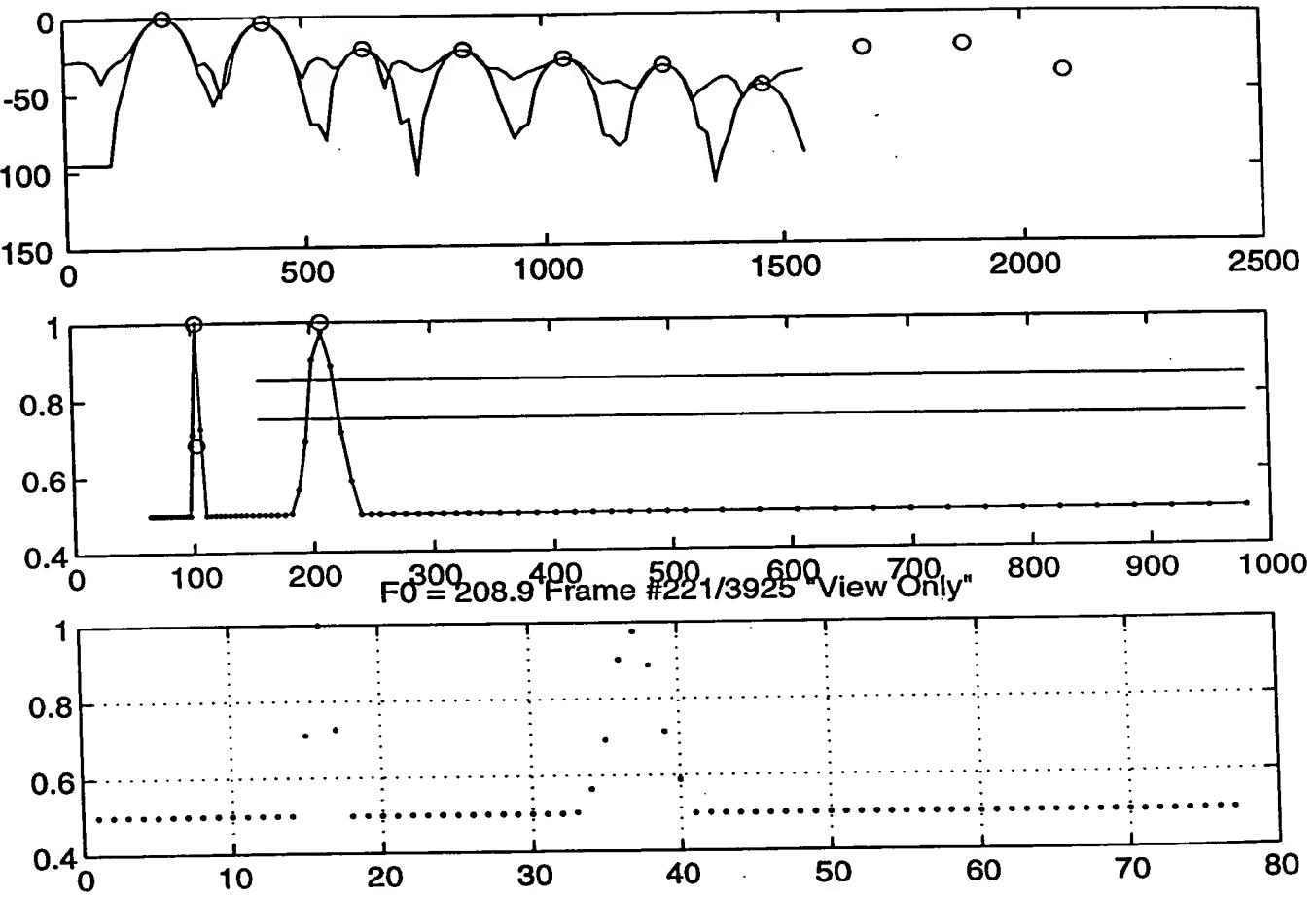


Fig. 9A

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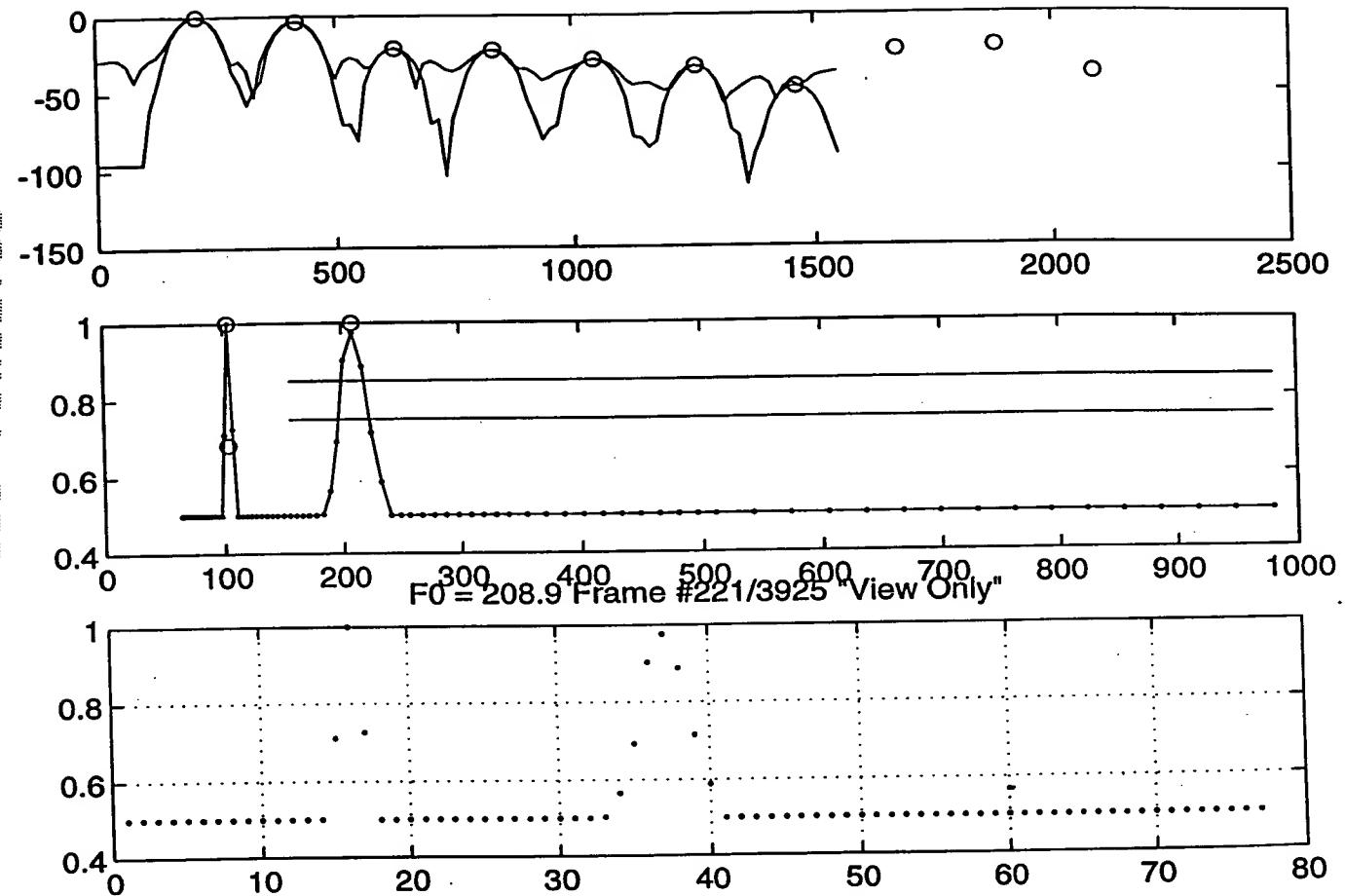


Fig. 9B

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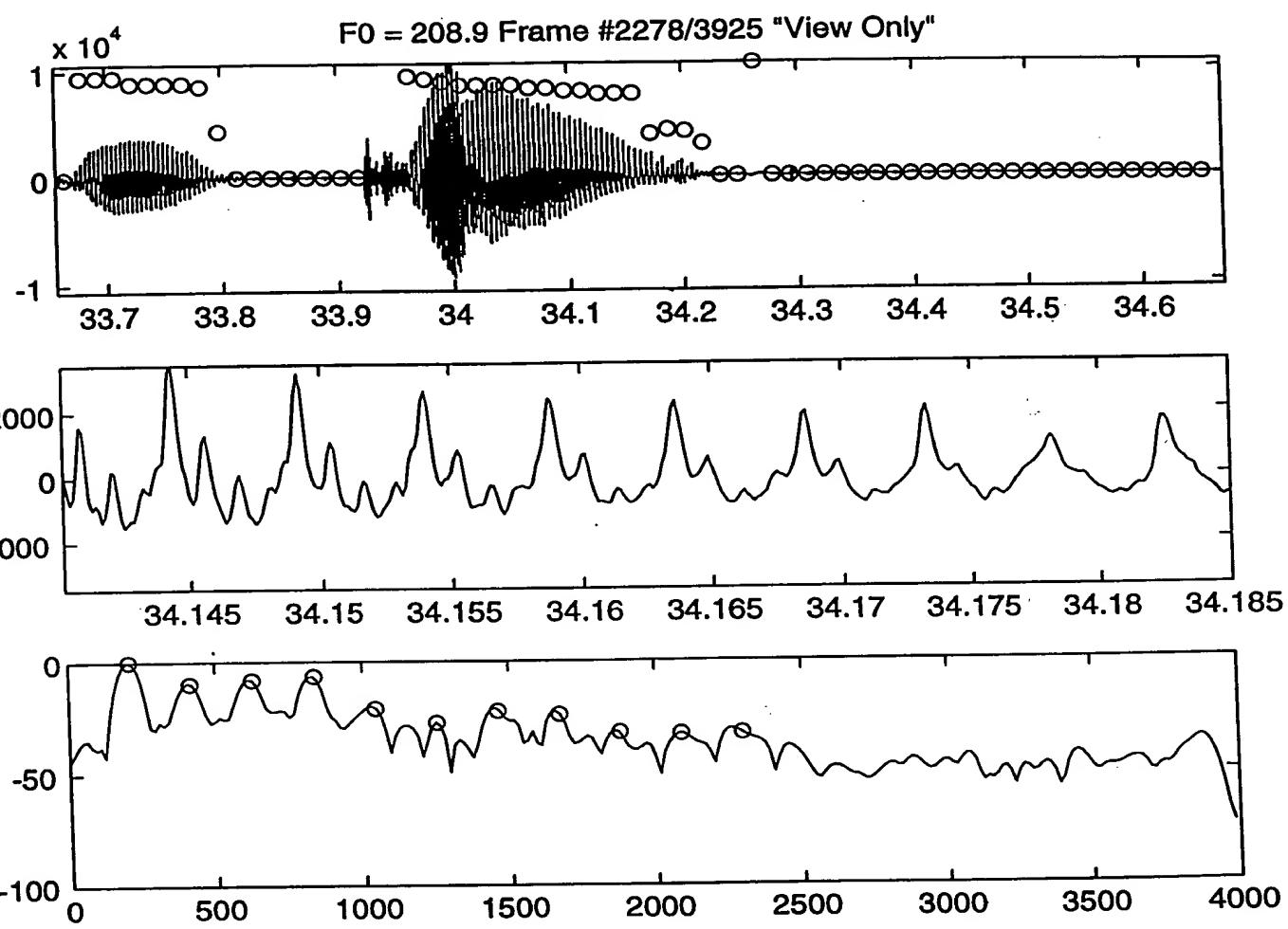


Fig. 9C

Sequence #2278/3925 View Only

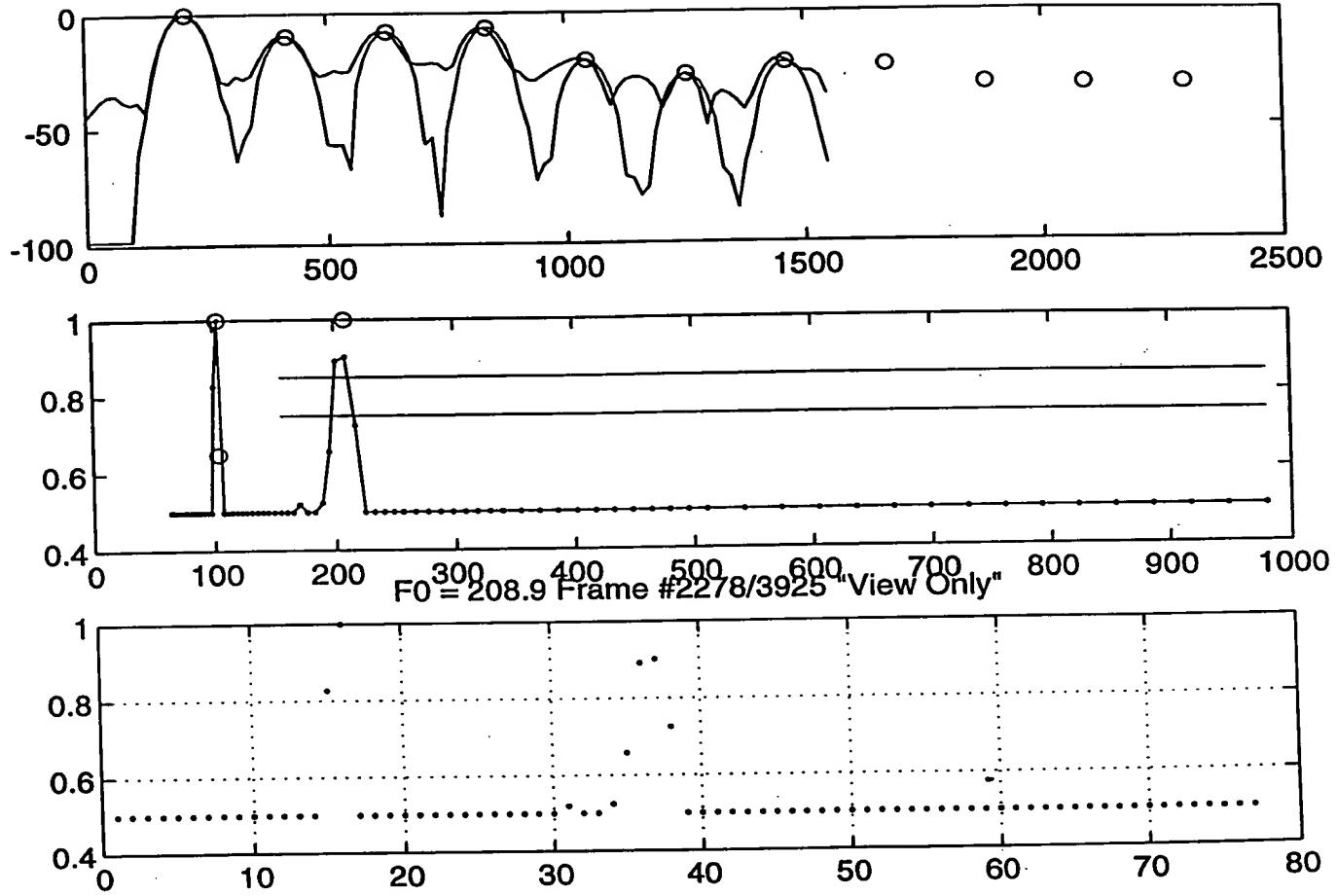


Fig. 9D

Fig. 10

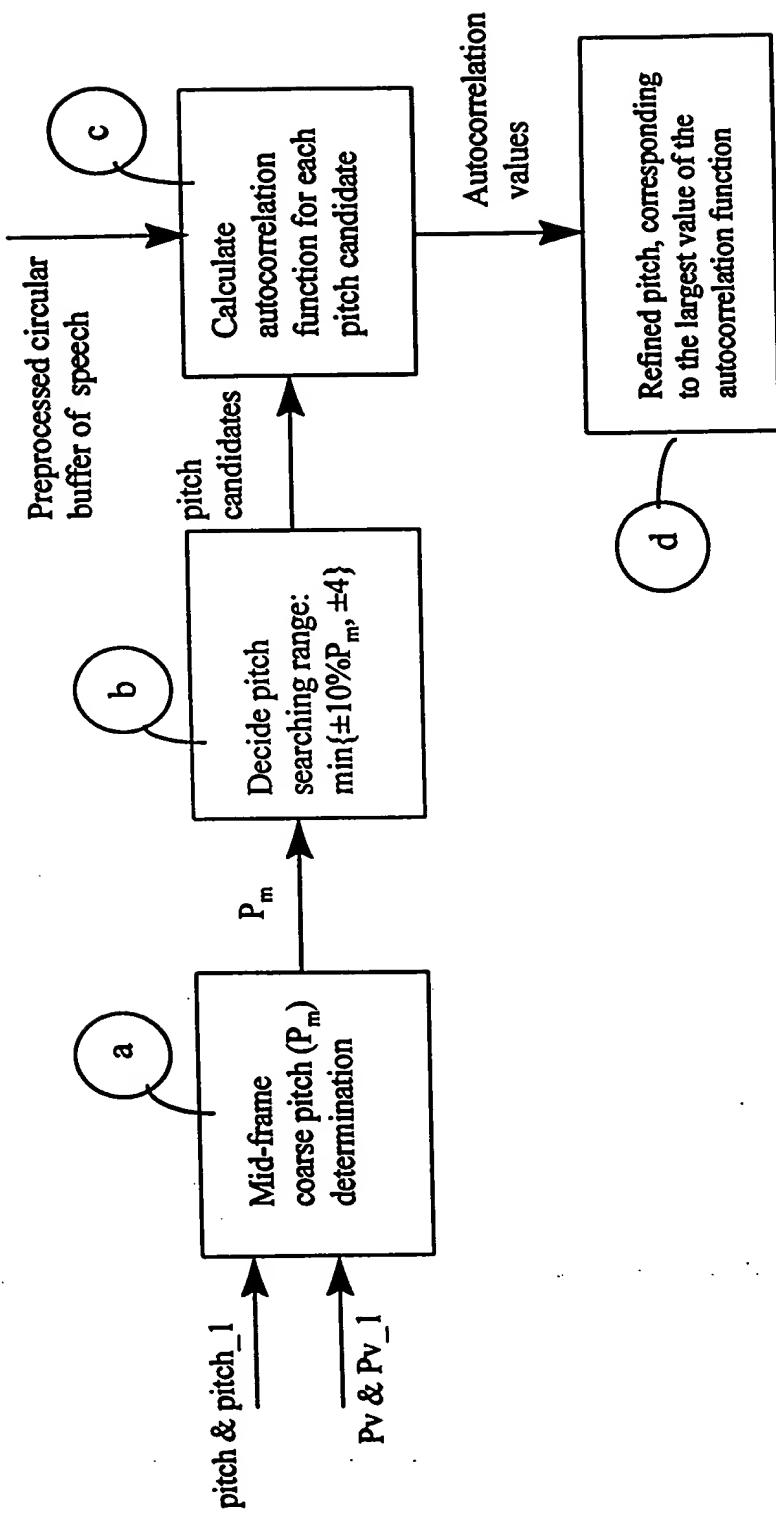


Fig. 11

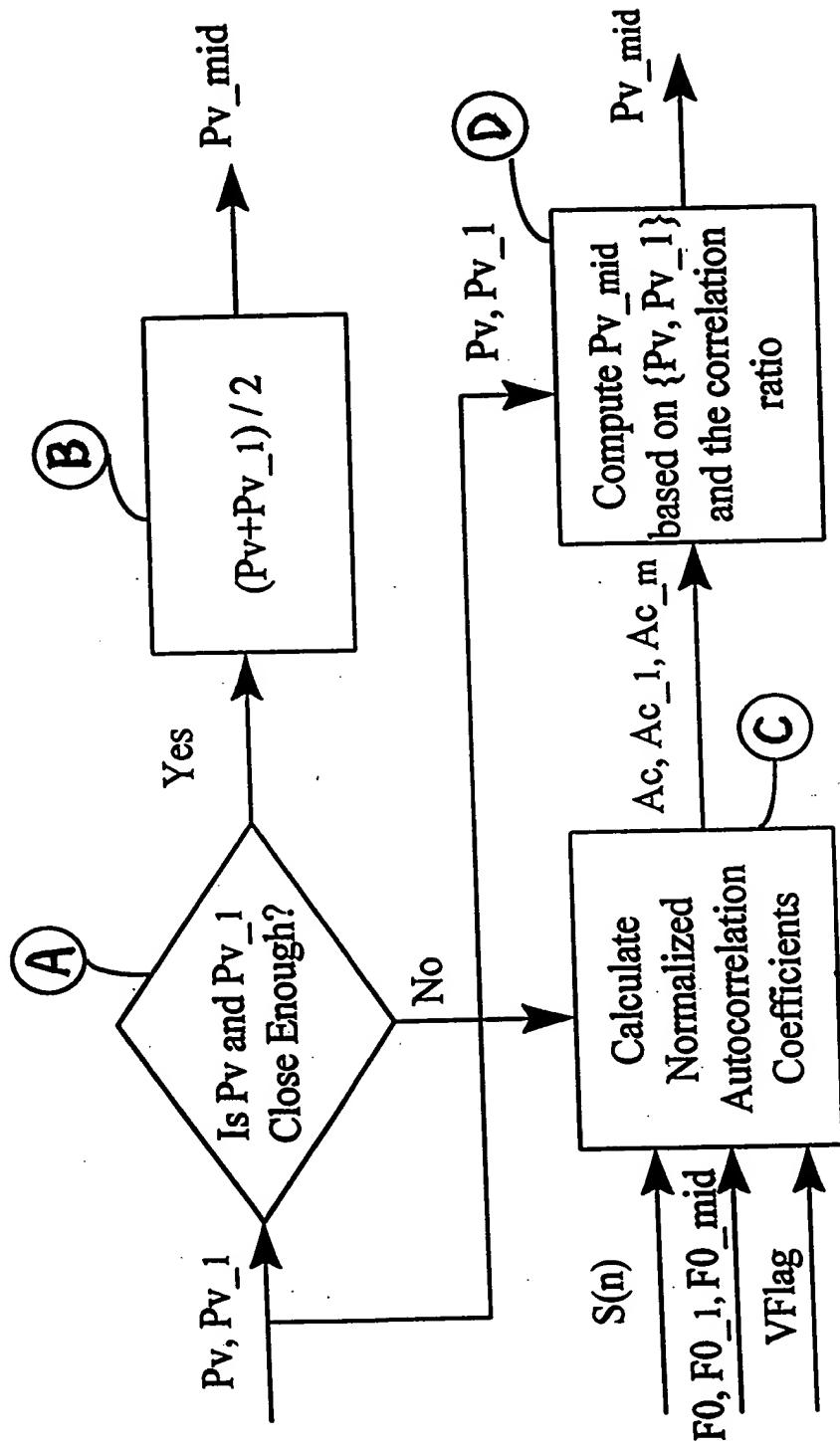


Fig. 12

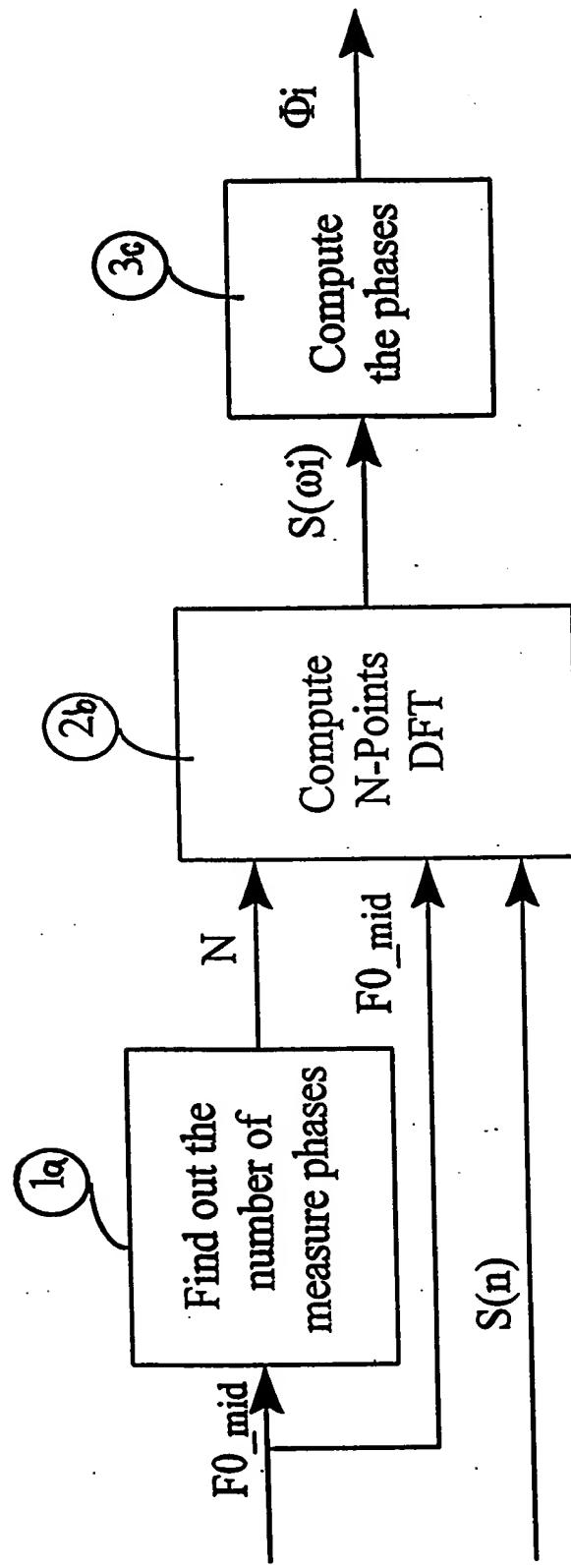
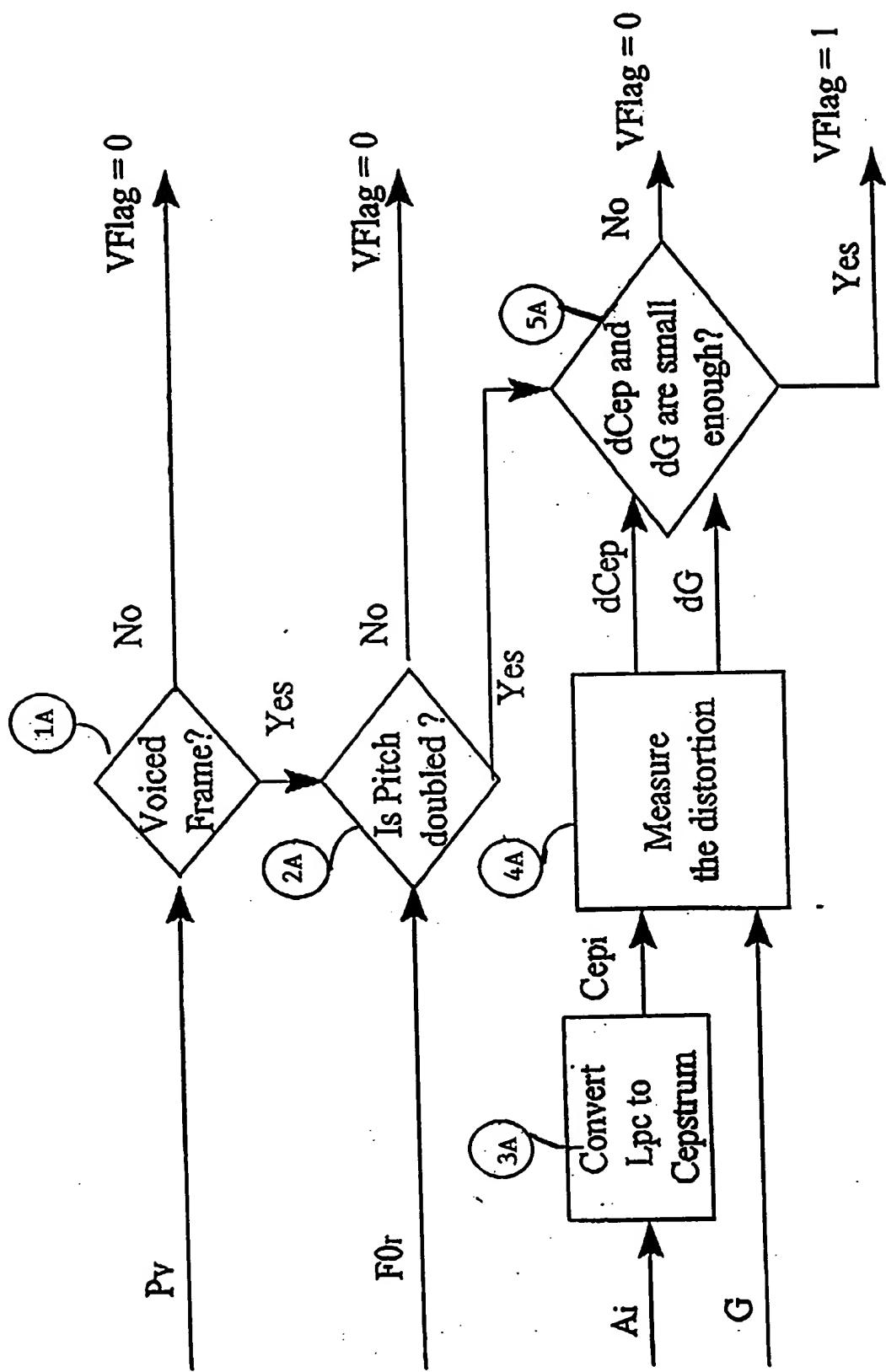


Fig. 13



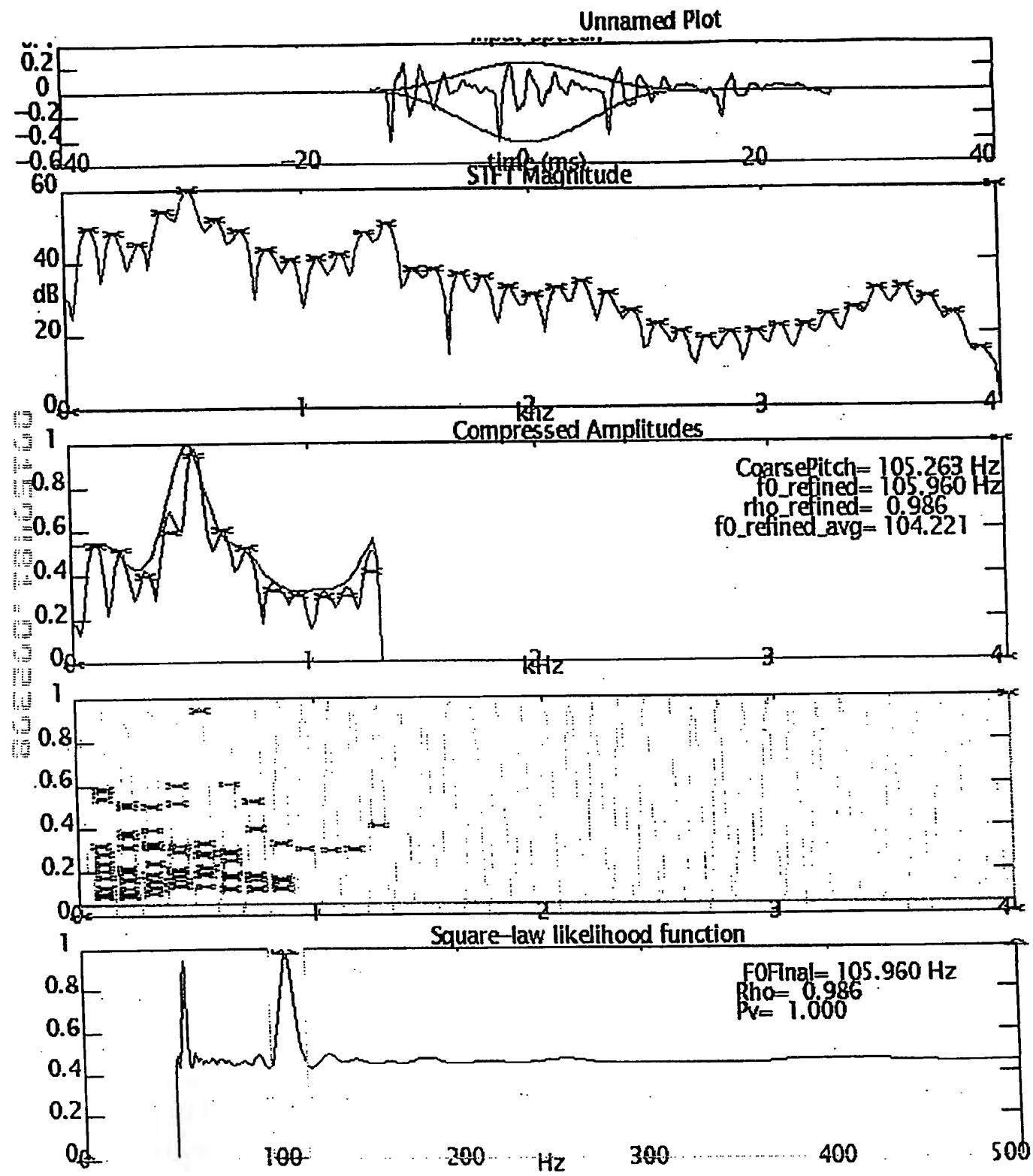
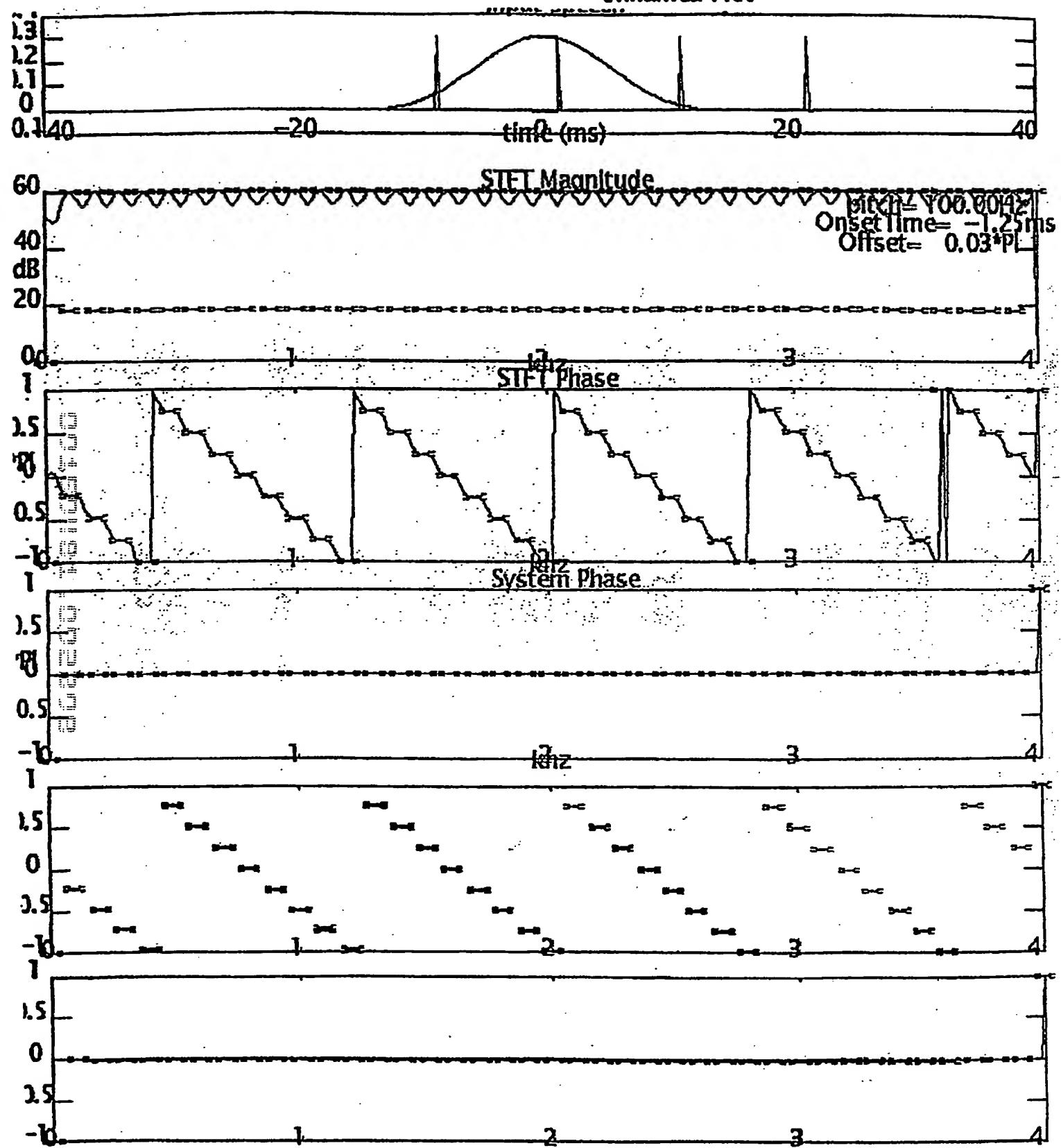
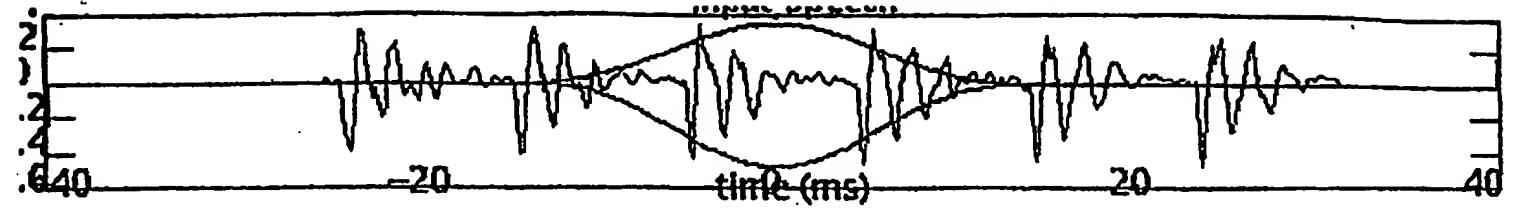


Fig. 14

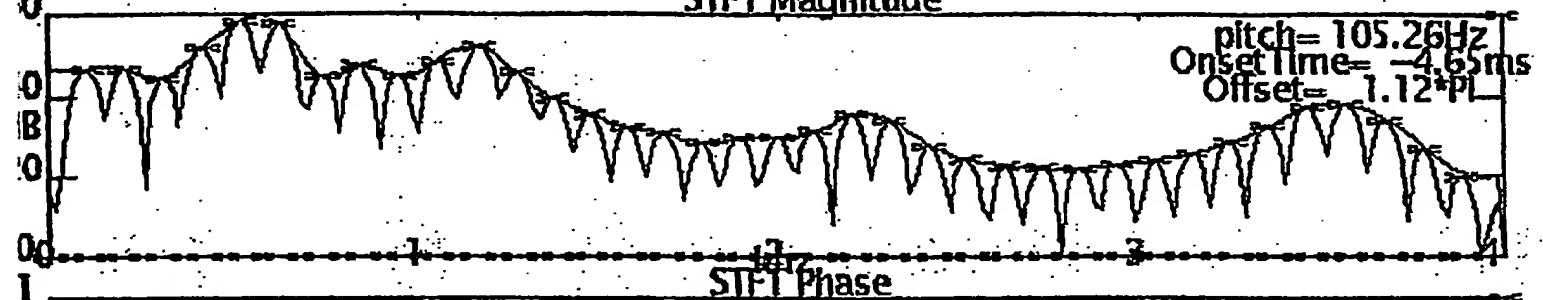
Unnamed Plot



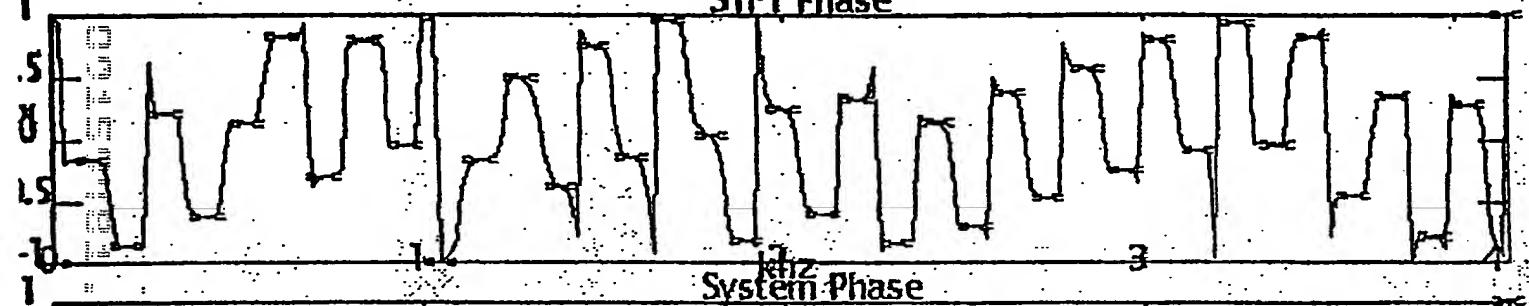
Unnamed Plot



STFT Magnitude



STFT Phase



System Phase

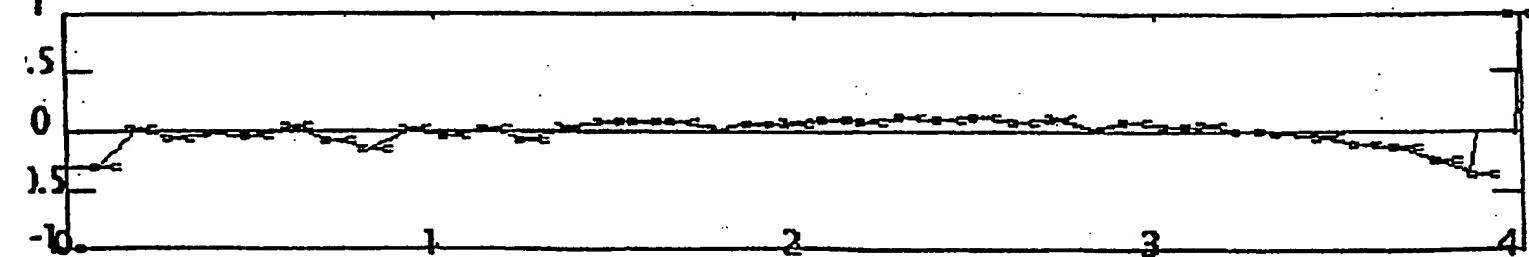
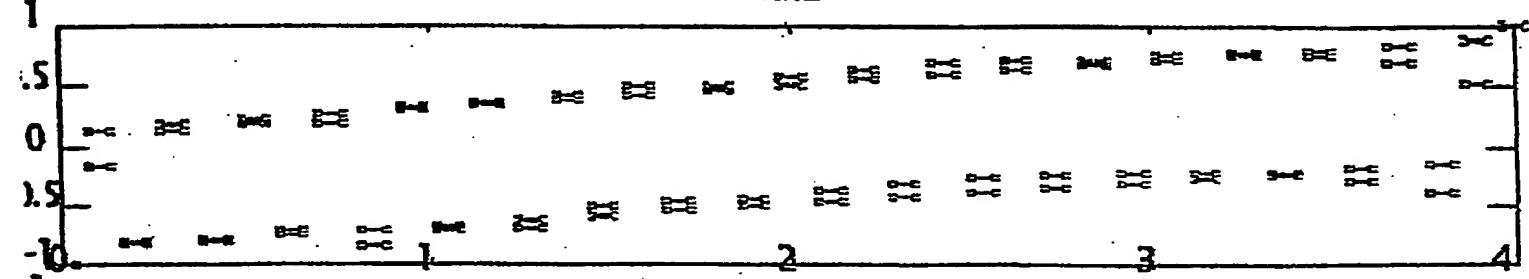
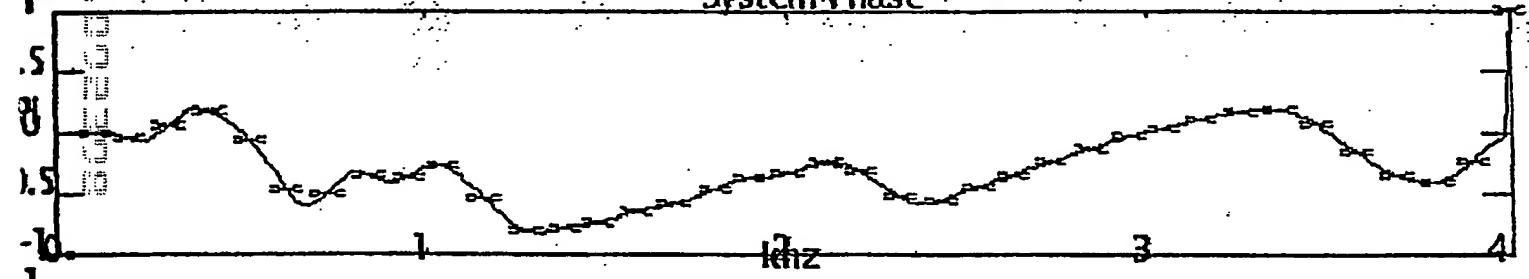
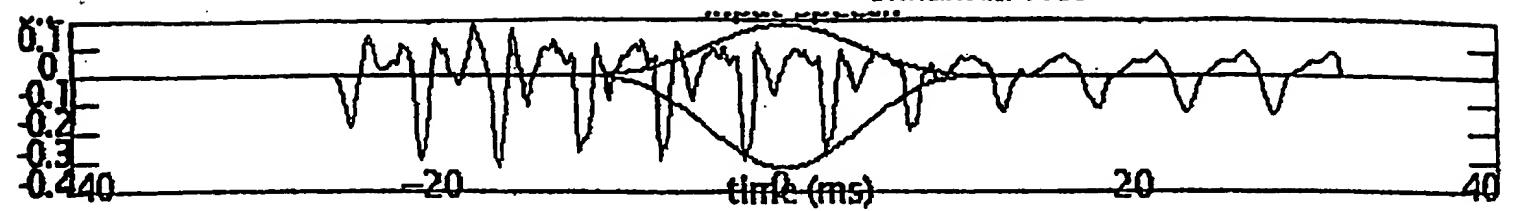
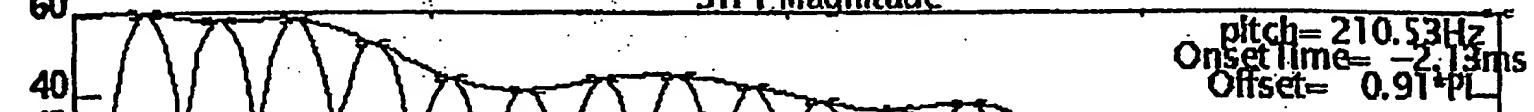


Fig. 16

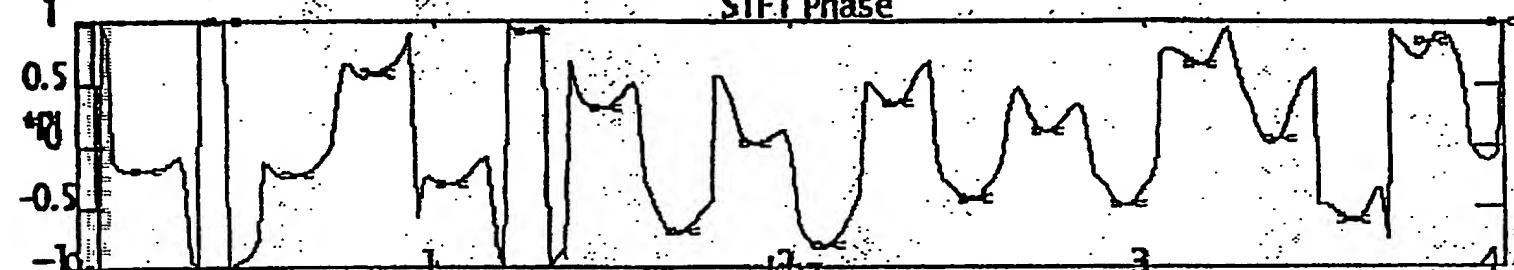
Unnamed Plot



STFT Magnitude



STFT Phase



System Phase

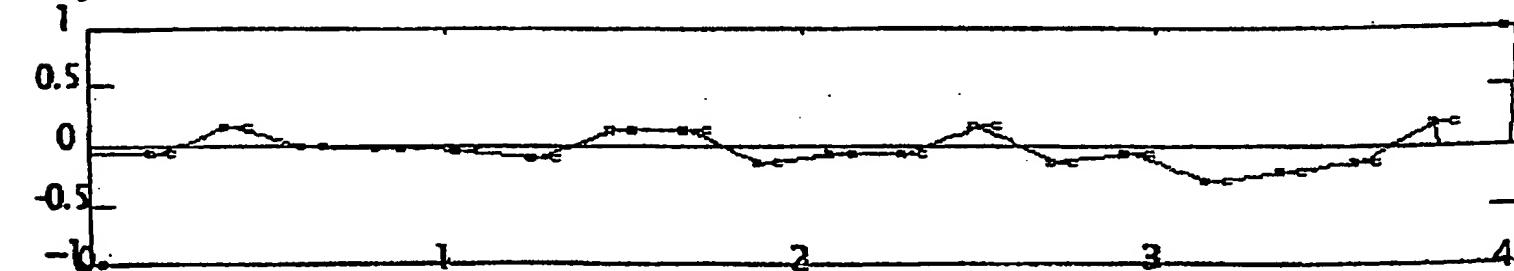
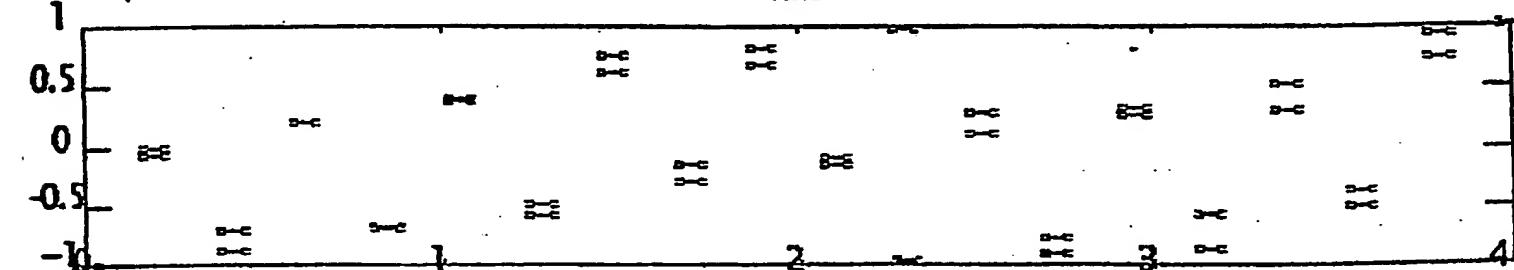
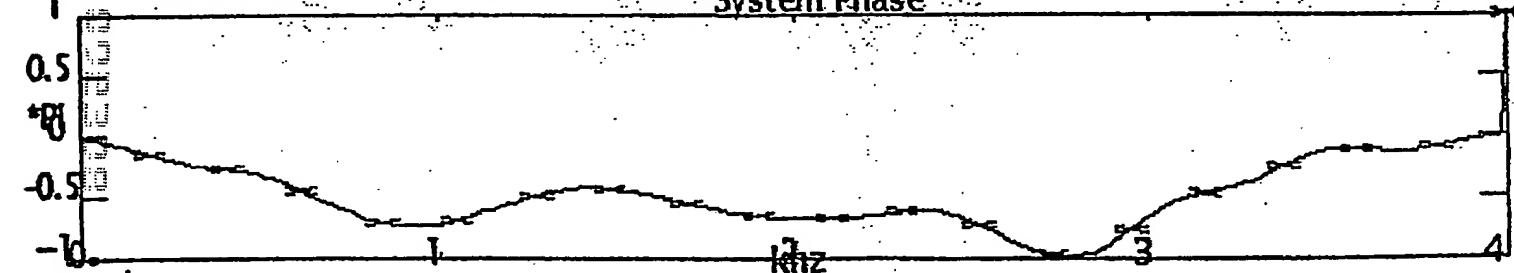


Fig. 17

Unnamed Plot

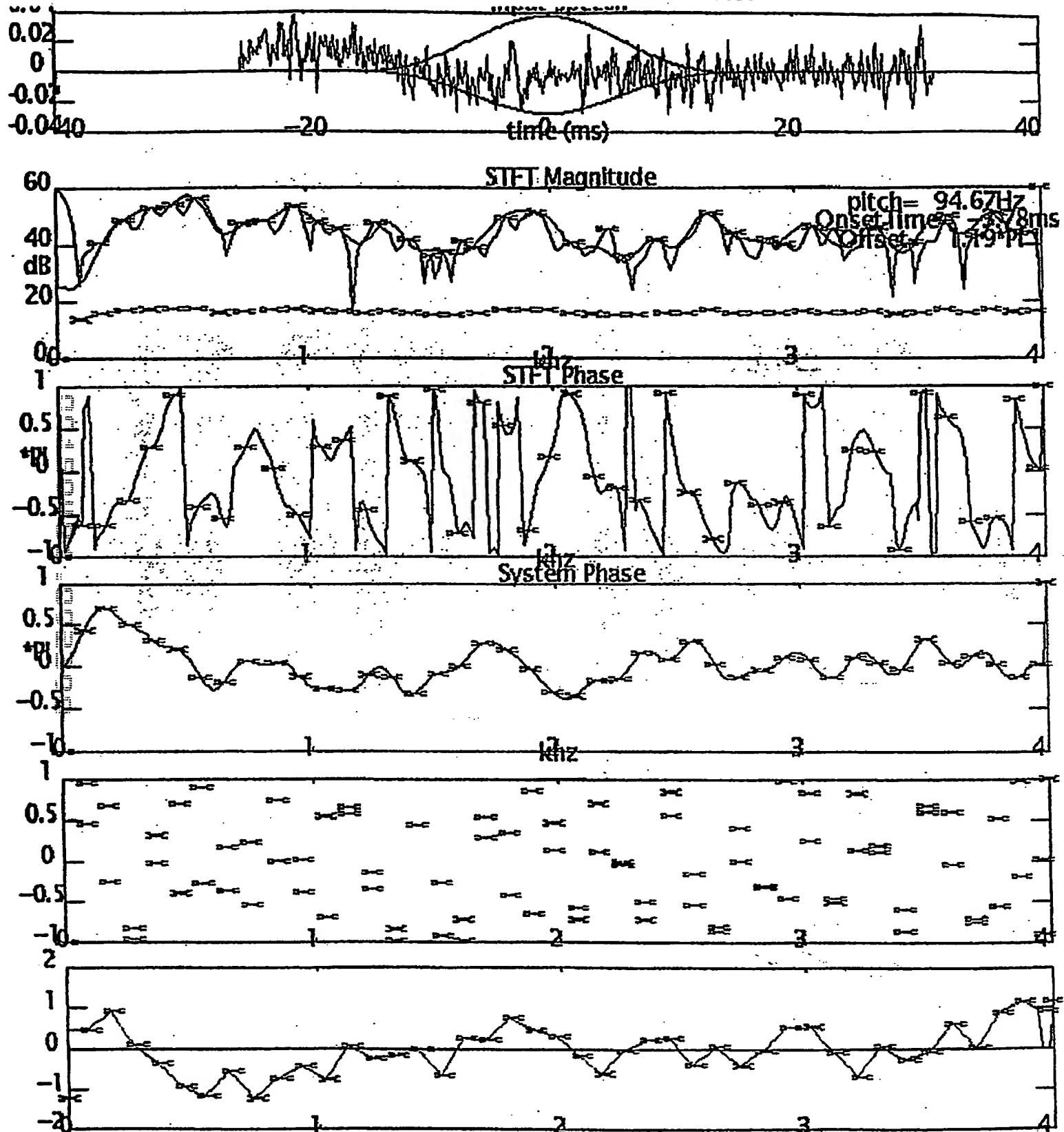
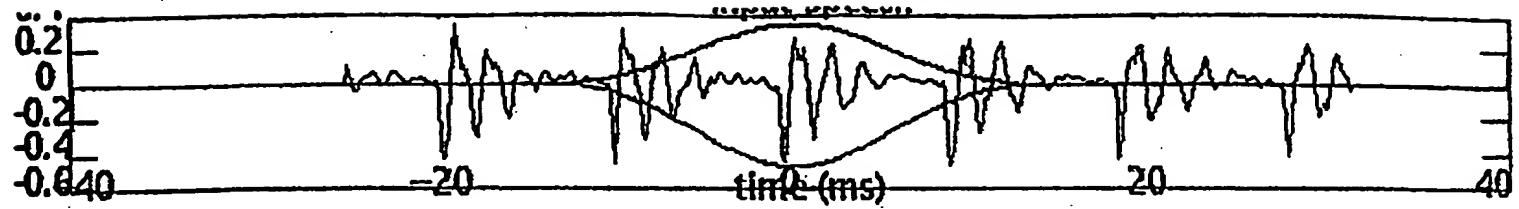
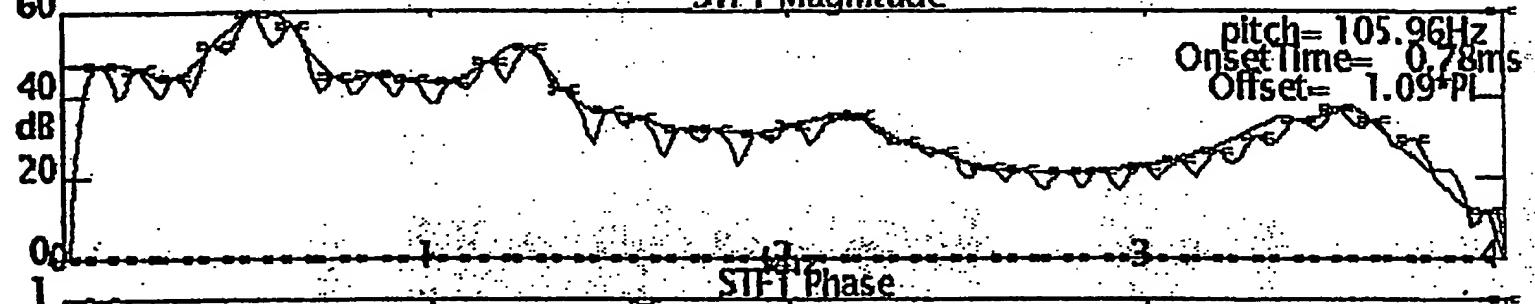


Fig. 18

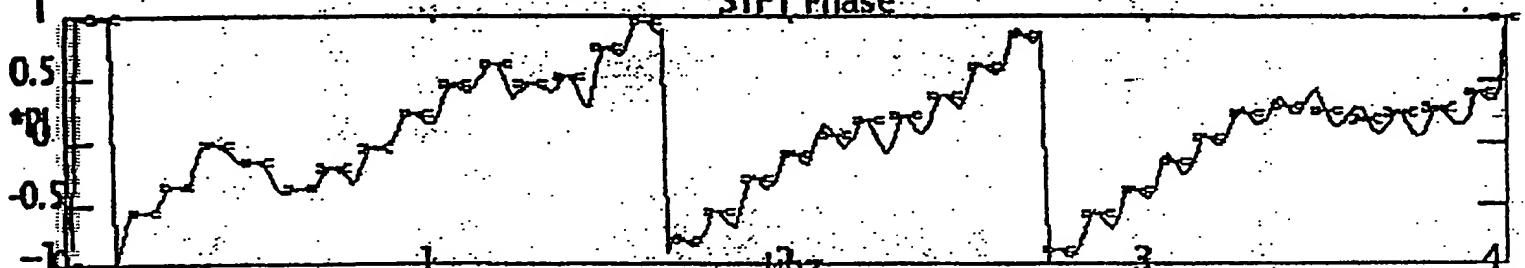
Unnamed Plot



STFT Magnitude



STFT Phase



System Phase

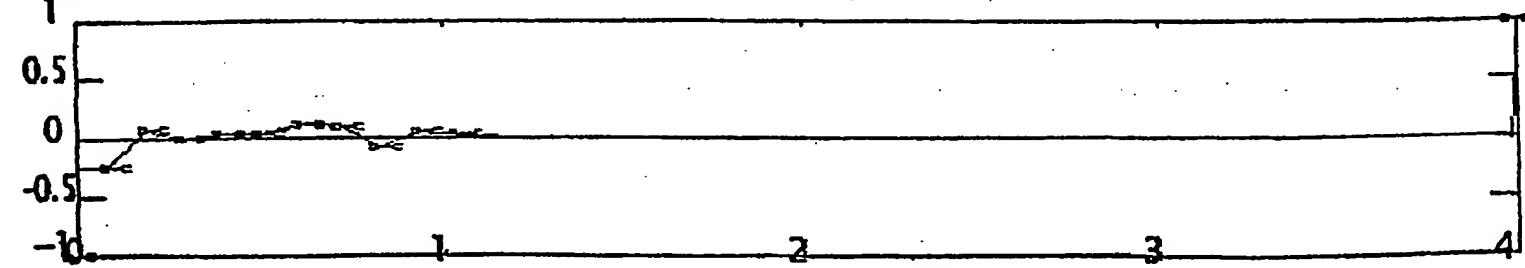
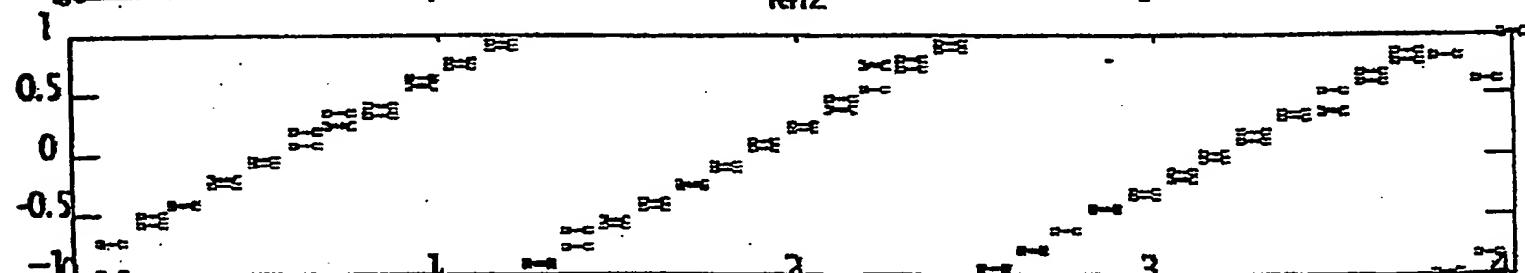
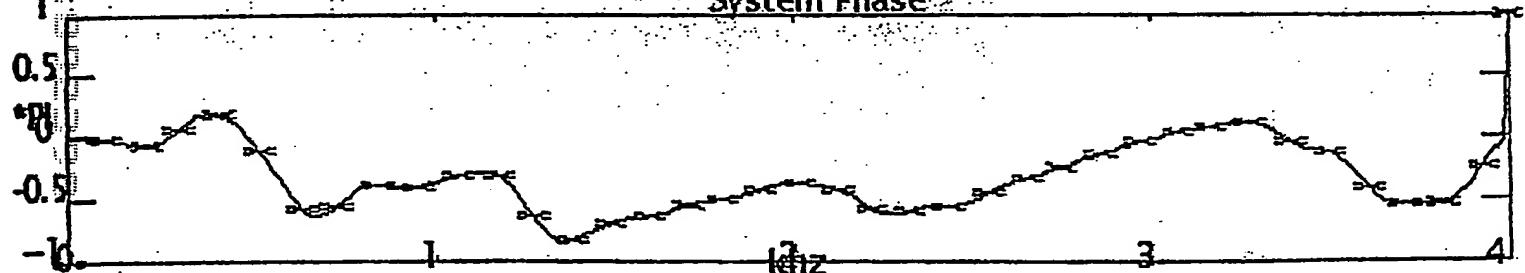


Fig. 19

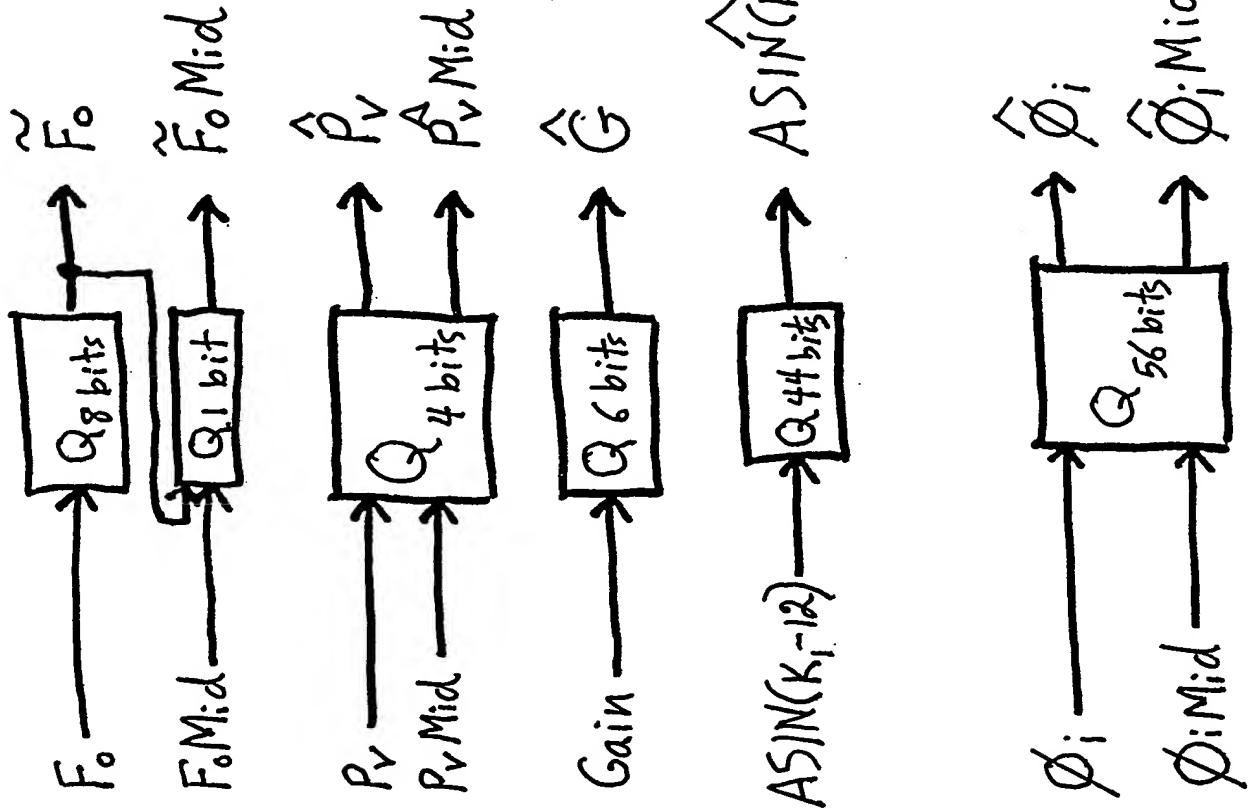
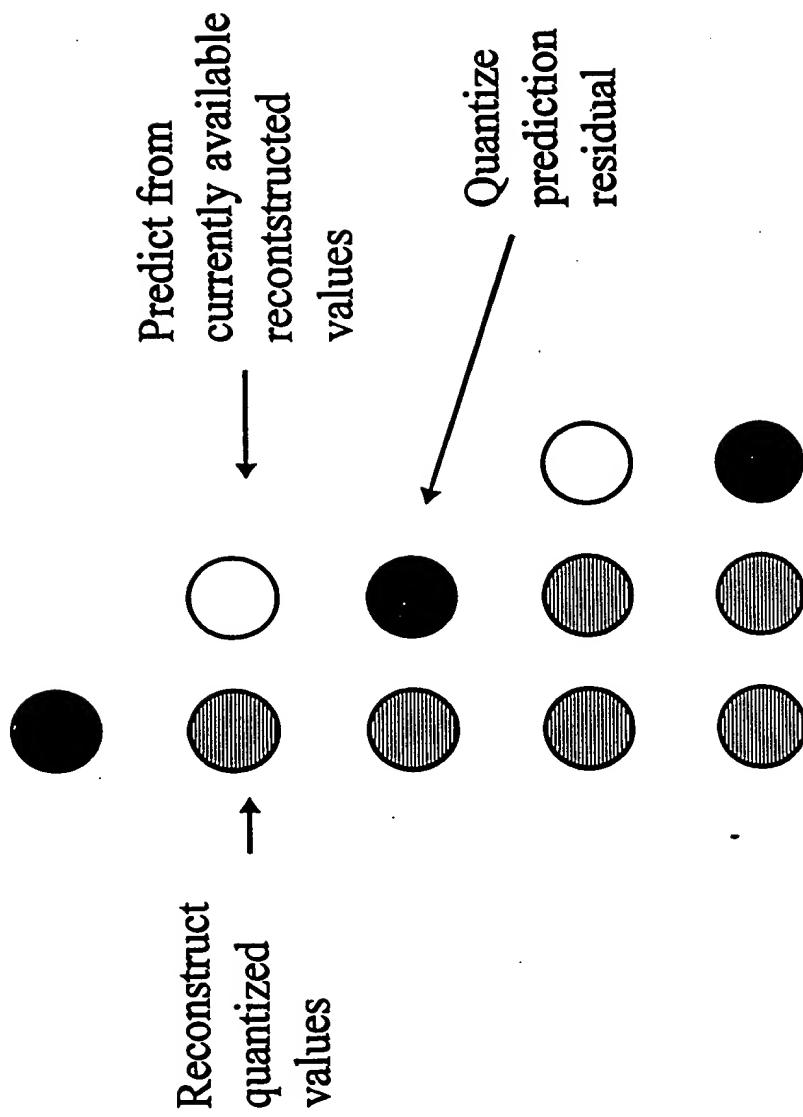
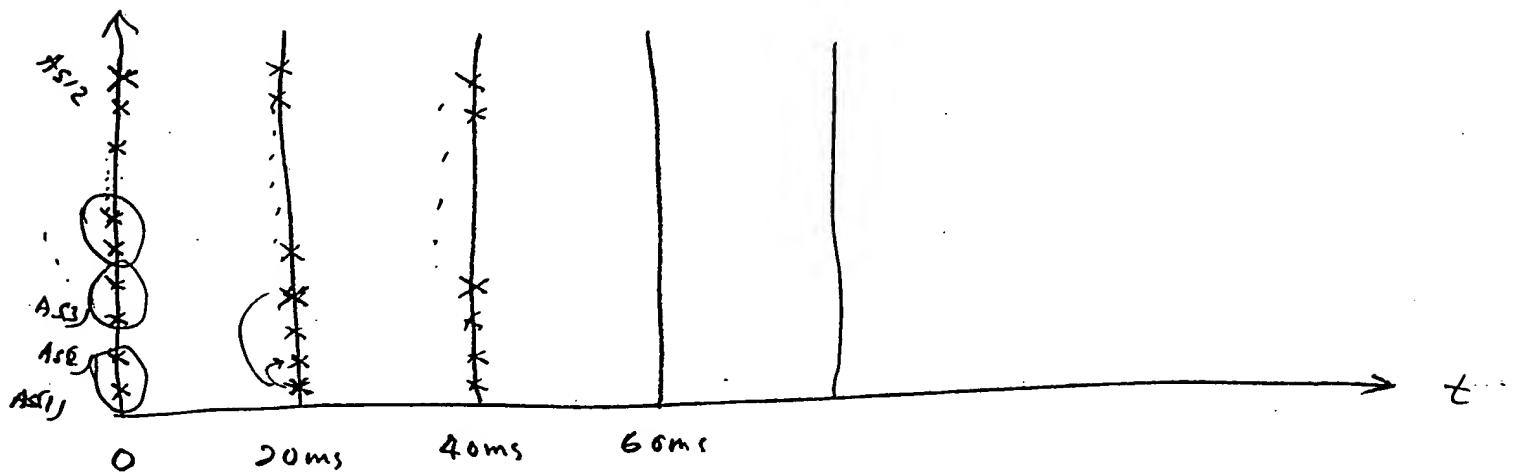


Fig. 20

Fig. 21



Spectral Quantization



prediction of $AS(i)$

$$\widehat{AS}(i) = \sum_{j=1}^{i-1} a_{ij} \tilde{AS}(j)$$

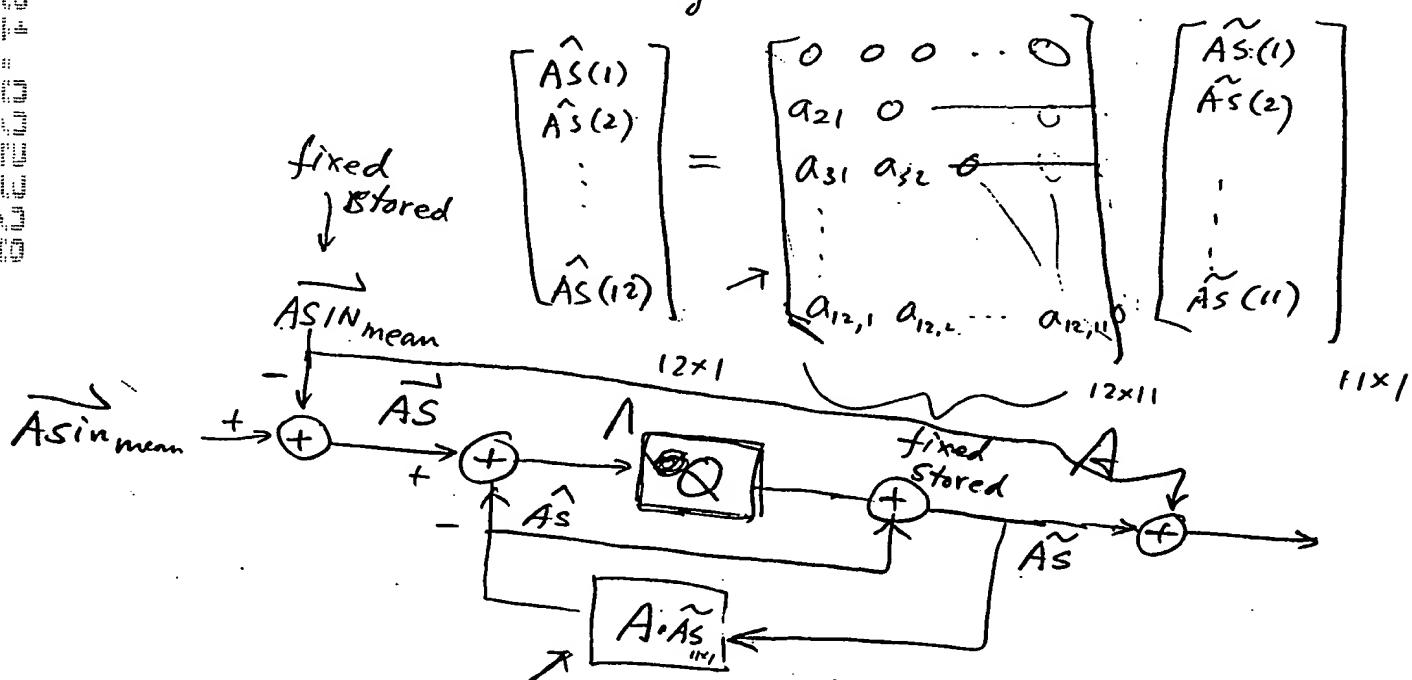


Fig. 21A

PHASE-PREDICTIVE CODING

w

w_{-1}

$$\bar{\theta} = \theta_{-1} + \frac{w_{-1} + w}{2} \times T$$

θ_{-1}

$$\text{PHASE RESIDUAL} = \theta - \bar{\theta}$$

w_{-1} = frequency at previous frame

w = frequency at current frame

θ_{-1} = quantized phase at previous frame

$\bar{\theta}$ = predicted phase at current frame

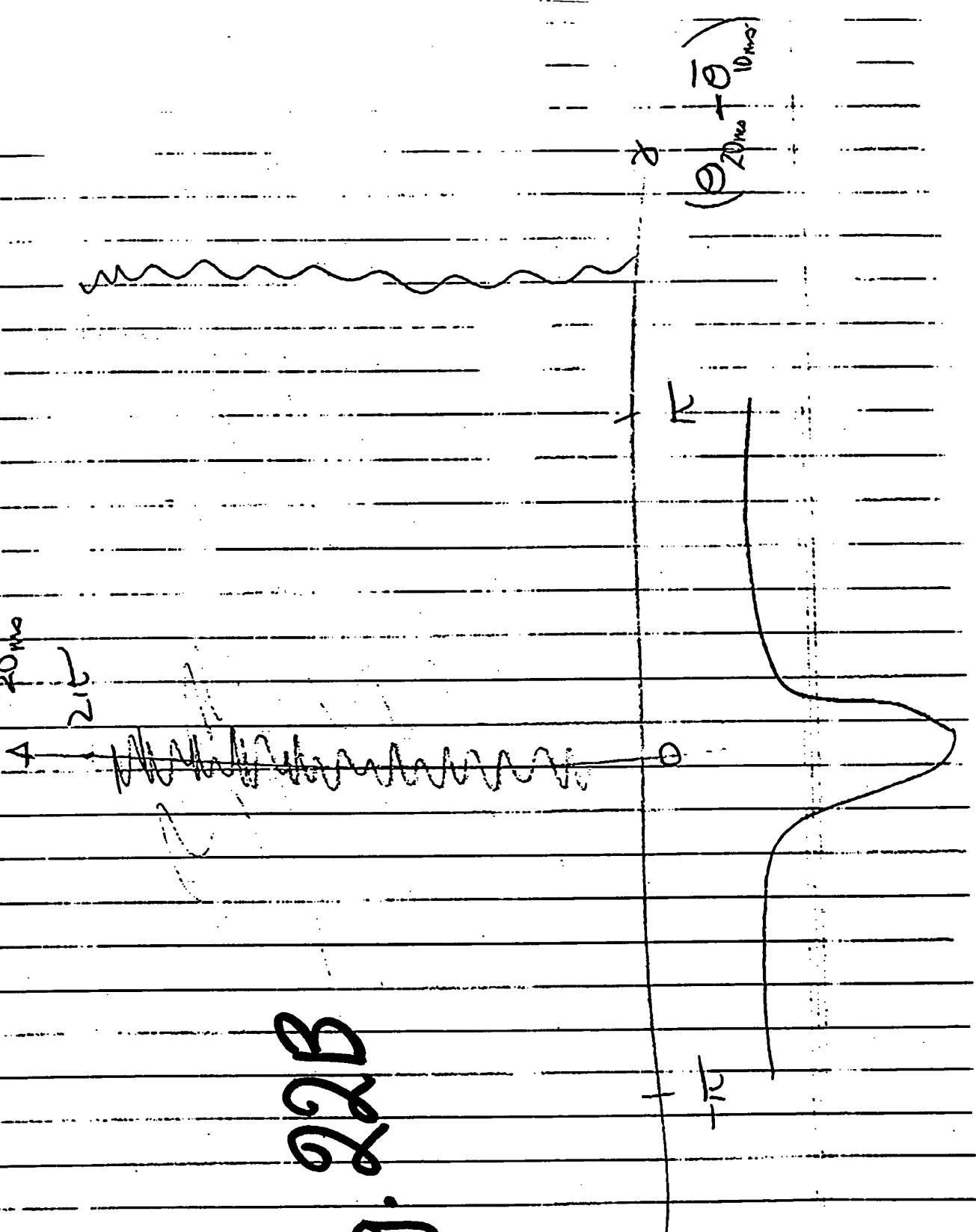
θ = measured phase at current frame

Fig. 22A

Scattering plot of prediction vs. $\theta_{20\text{ns}}$ AND $\theta_{10\text{ns}}$ PHASE

OFF ELECO
① 20 ns
② 10 ns

Fig. 22B



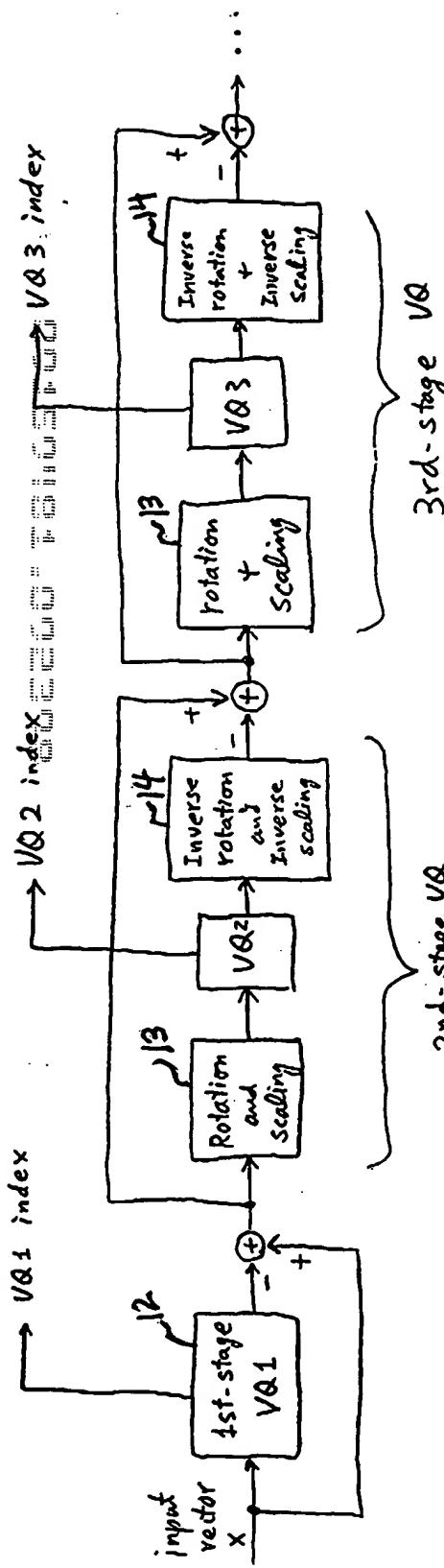


Fig. 23A

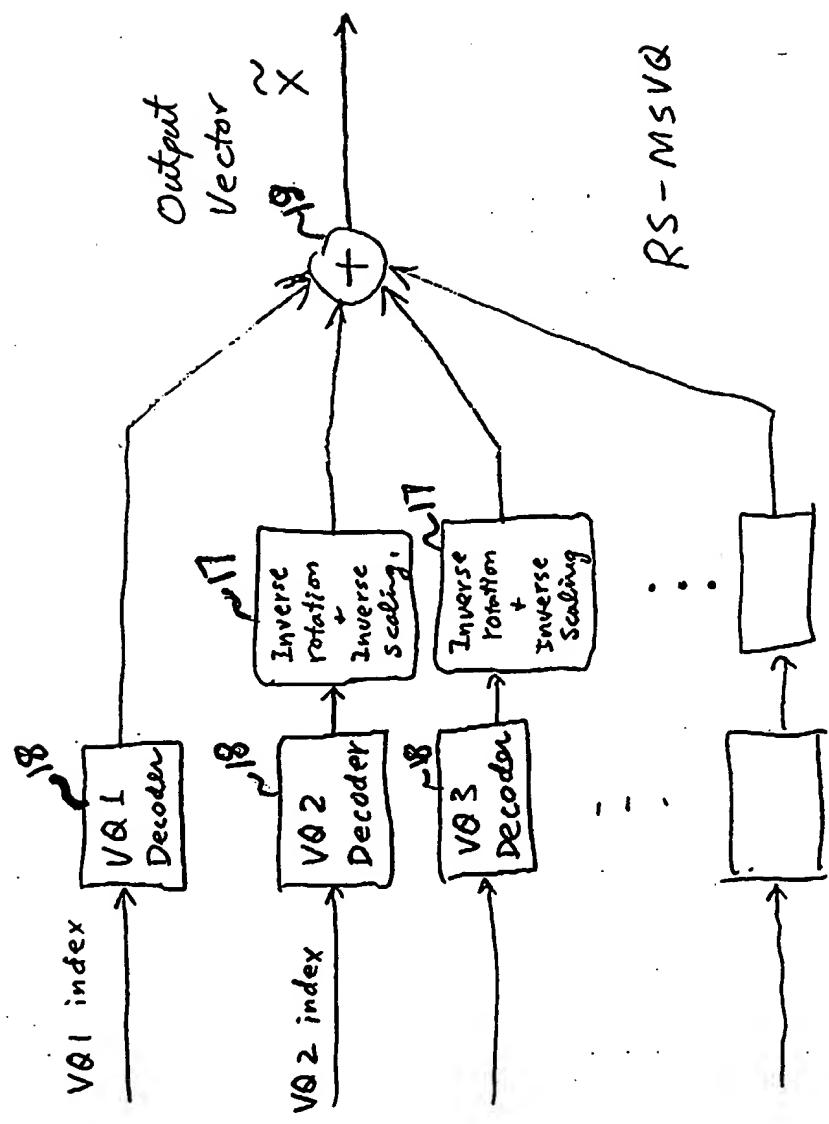


Fig. 23B

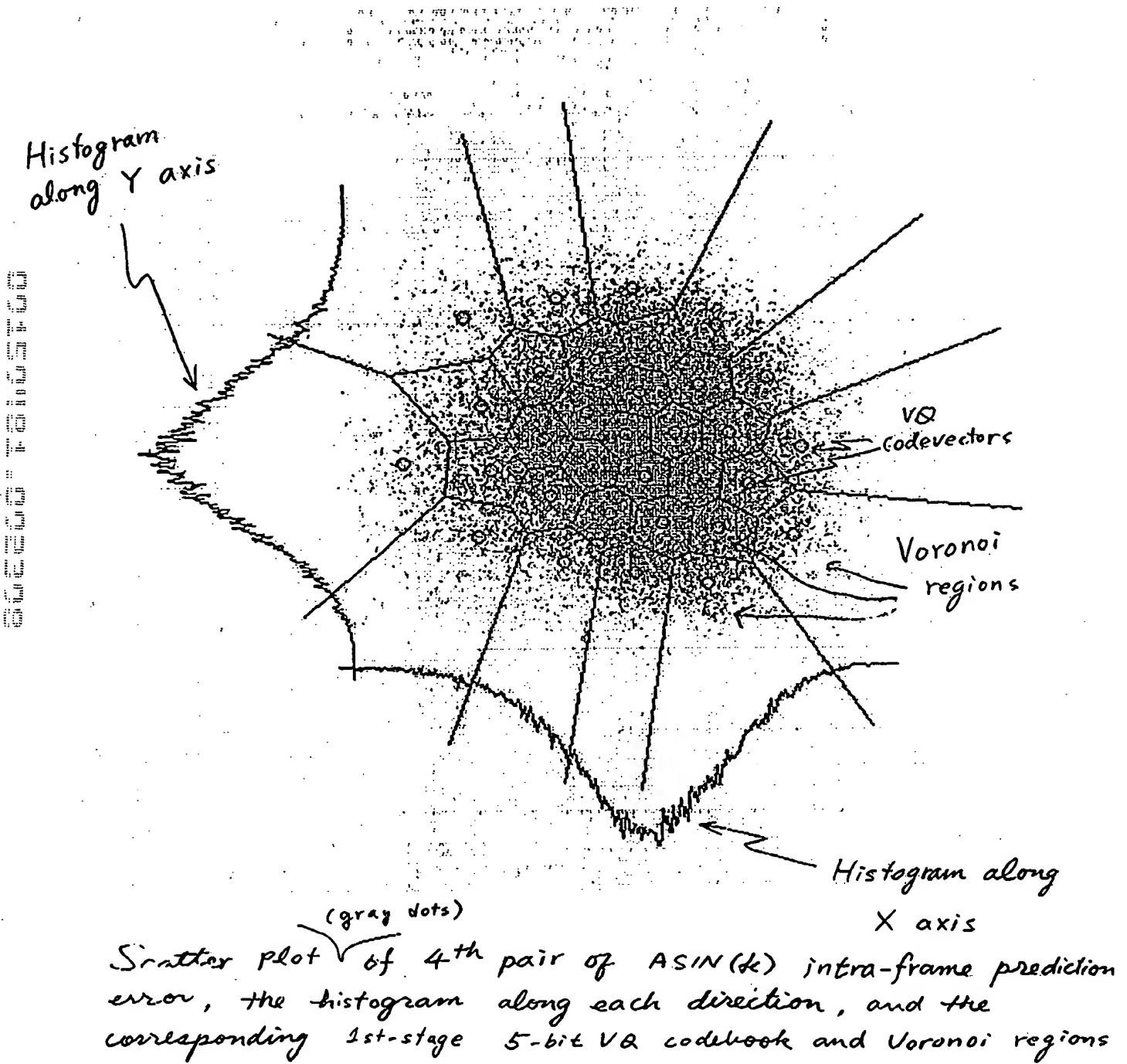


Fig. 24A

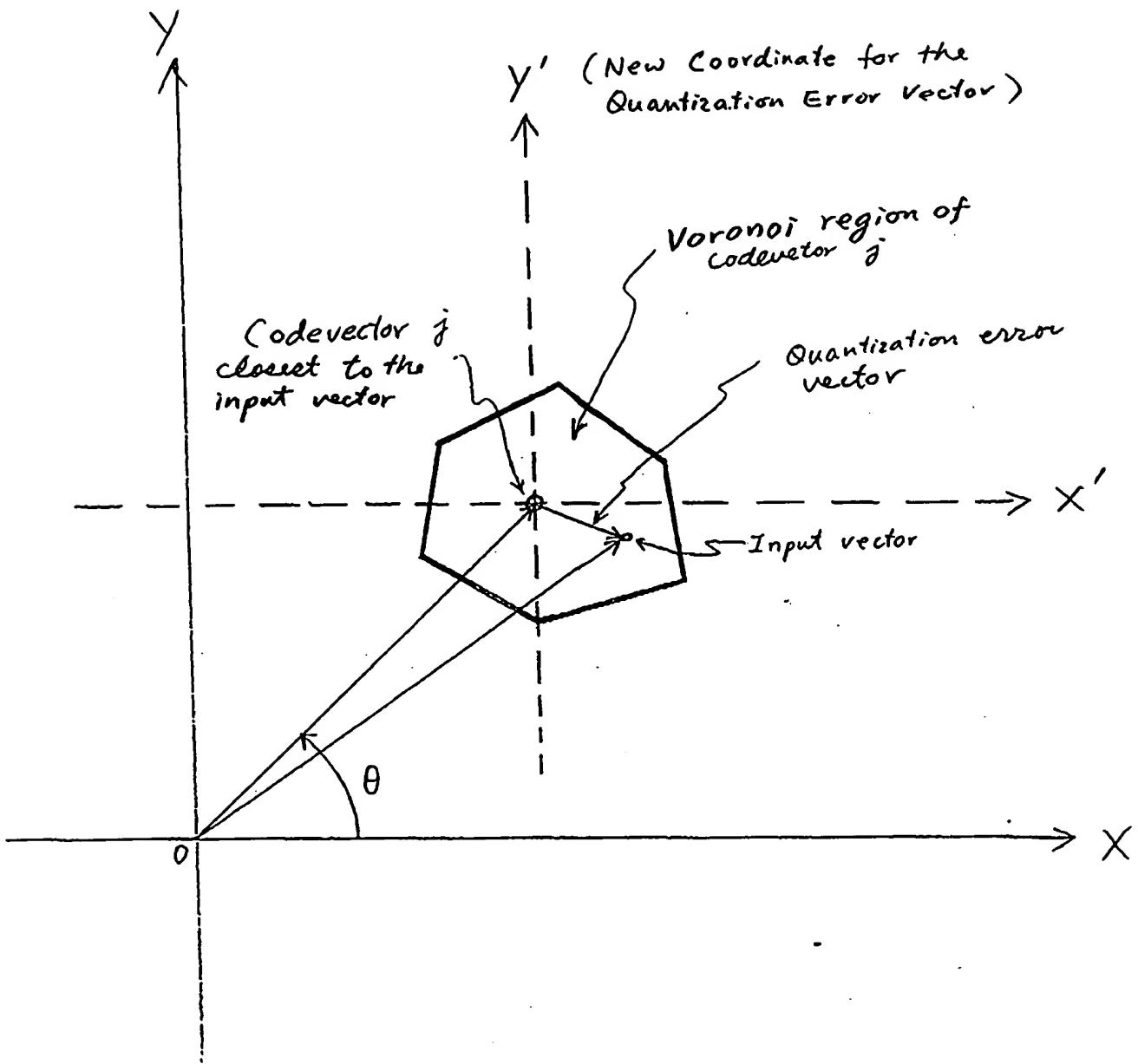
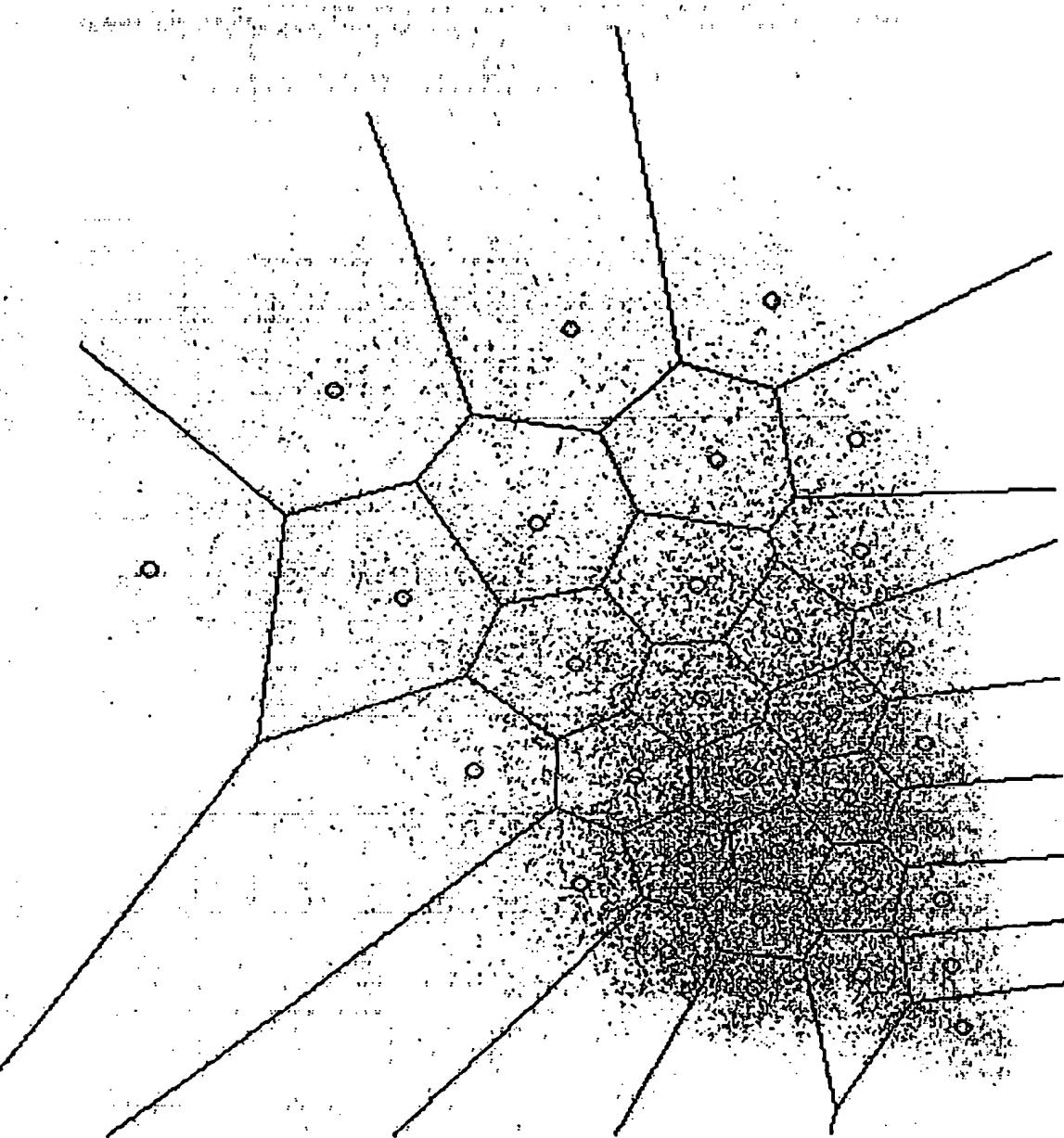


Fig. 24 B Illustration of the effect of subtracting the closest codevector for the input vector to get the quantization error vector.

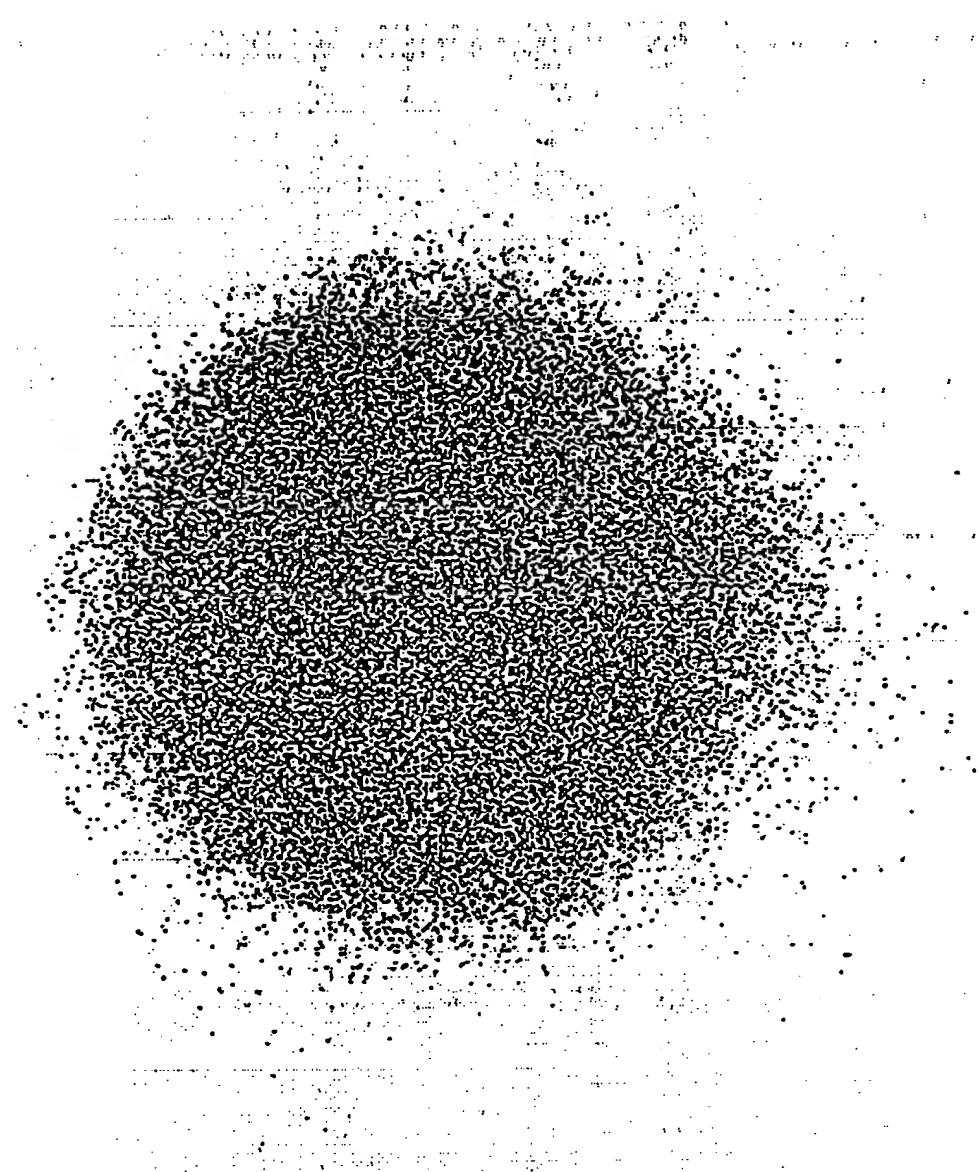
Fig. 24B

Scatter plot of 1st pair of ASIN(k) (gray dots) and
1st-stage VQ codebook (small circles), and the
corresponding Voronoi cells



Scatter plot of 1st pair of ASIN(k) (gray dots) and
1st-stage VQ codebook (small circles), and the
corresponding Voronoi cells

Fig. 24C

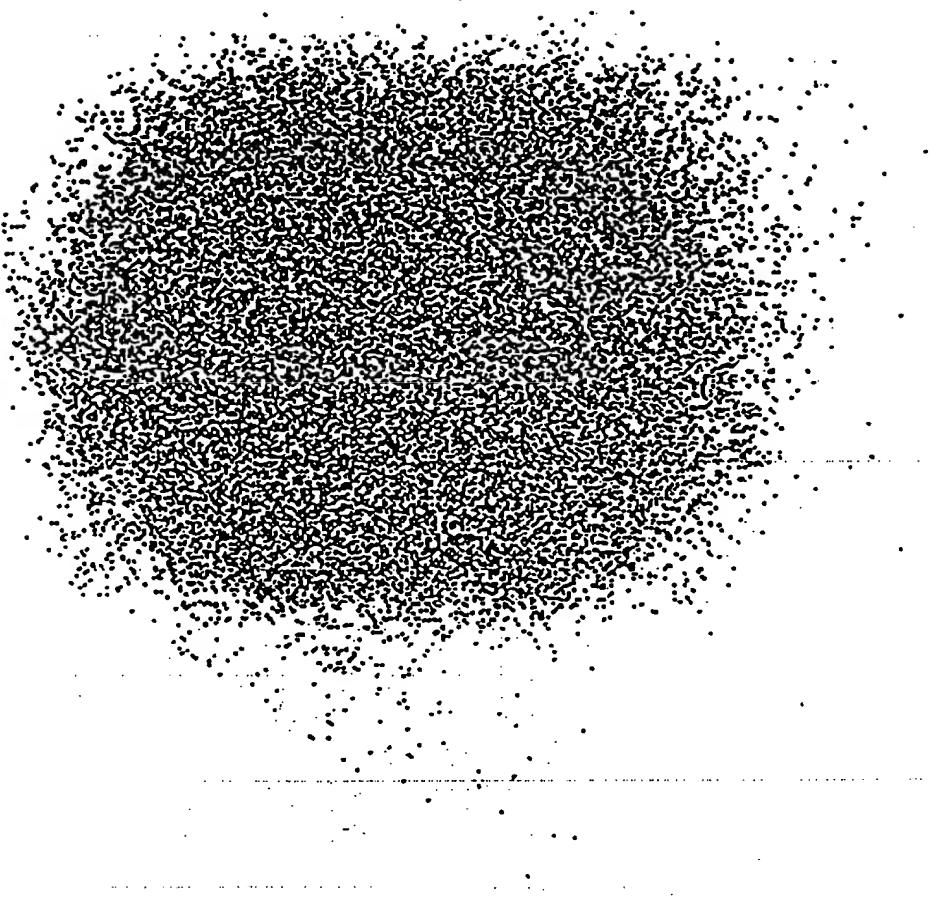


Without Hand-tuned Rotation Angles

— inner cells of 1st pair of ASIN(k)
1st-stage VQ of

Fig. 25

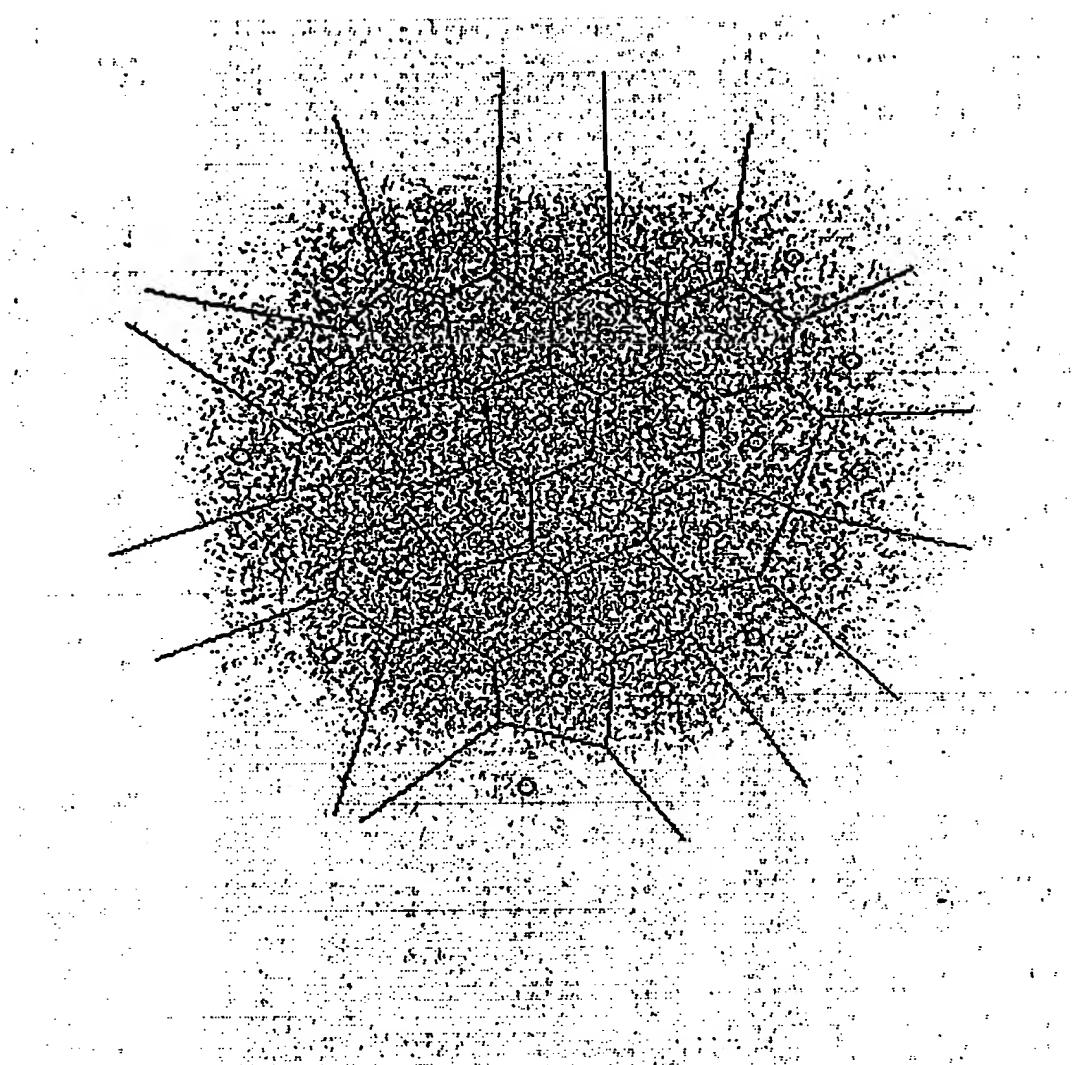
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With hand-tuned rotation angles
— inner cells of 1st pair of ASIN(k)
of 1st-stage VQ of

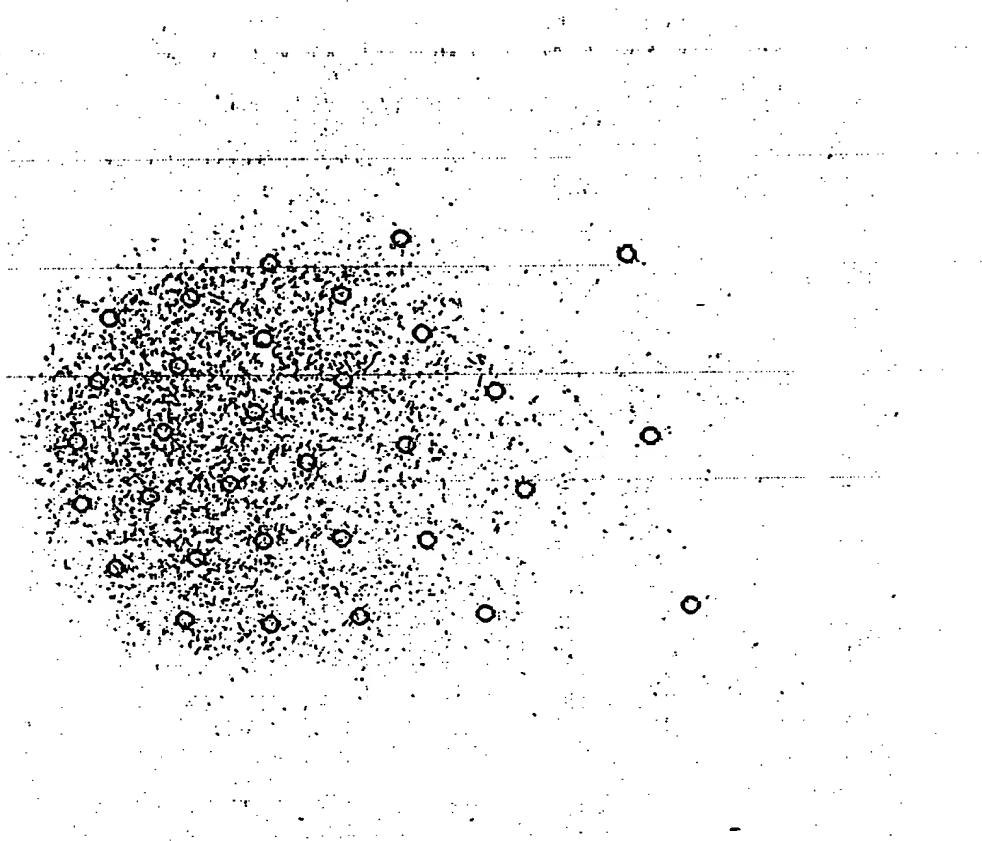
Fig. 26

0000000000000000



Inner-cell 1st-stage VQ error vector distribution (gray dots)
(hand tuning) and corresponding 2nd-stage VQ codebook (small circles)
for 1st pair of ASIN(k)

Fig. 27



Outer-cell 1st-stage VQ error vector distribution
and corresponding 2nd-stage VQ codebook (small circles)
for 1st pair of ASIN(k)

Fig. 28

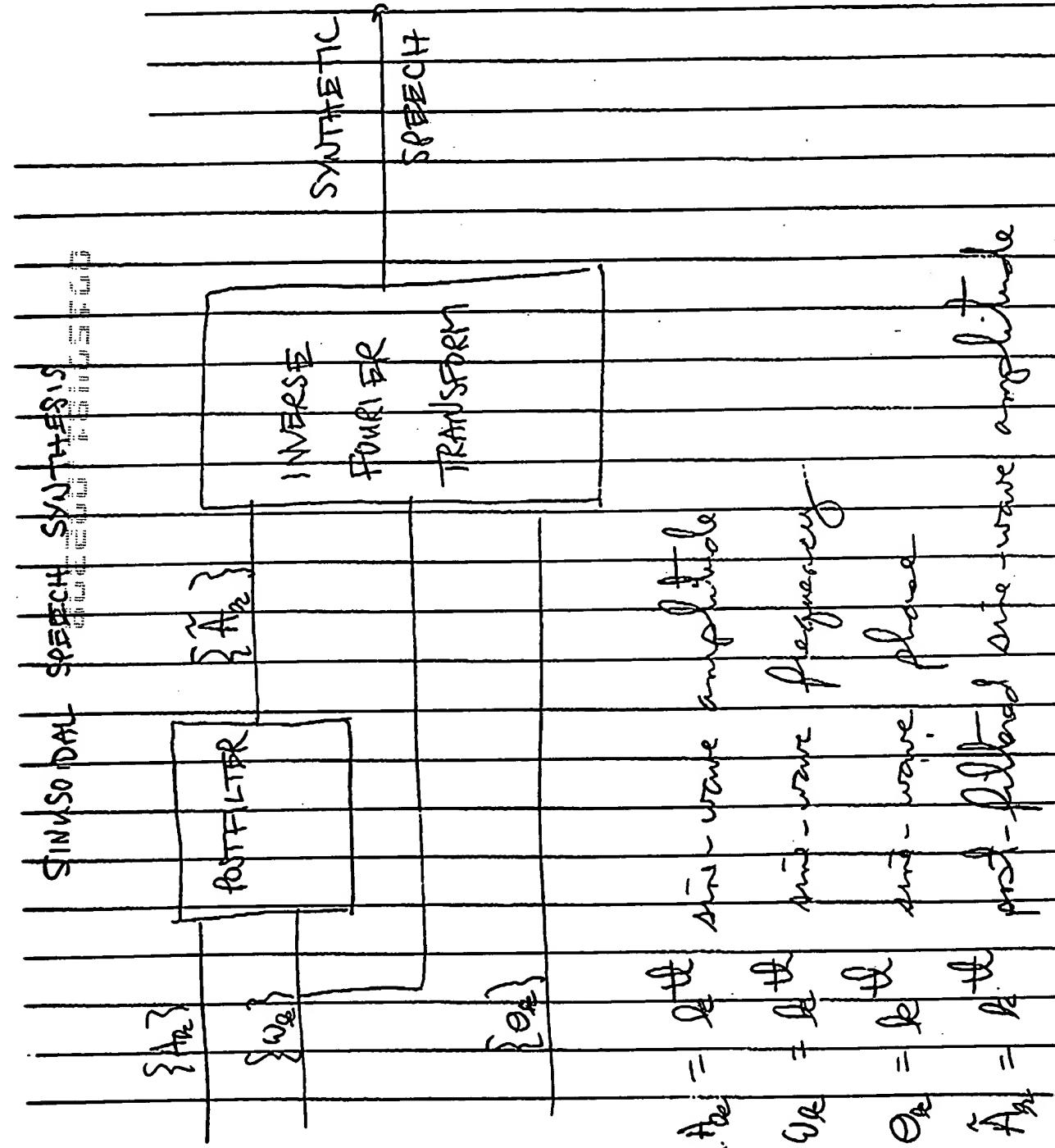


Fig. 29

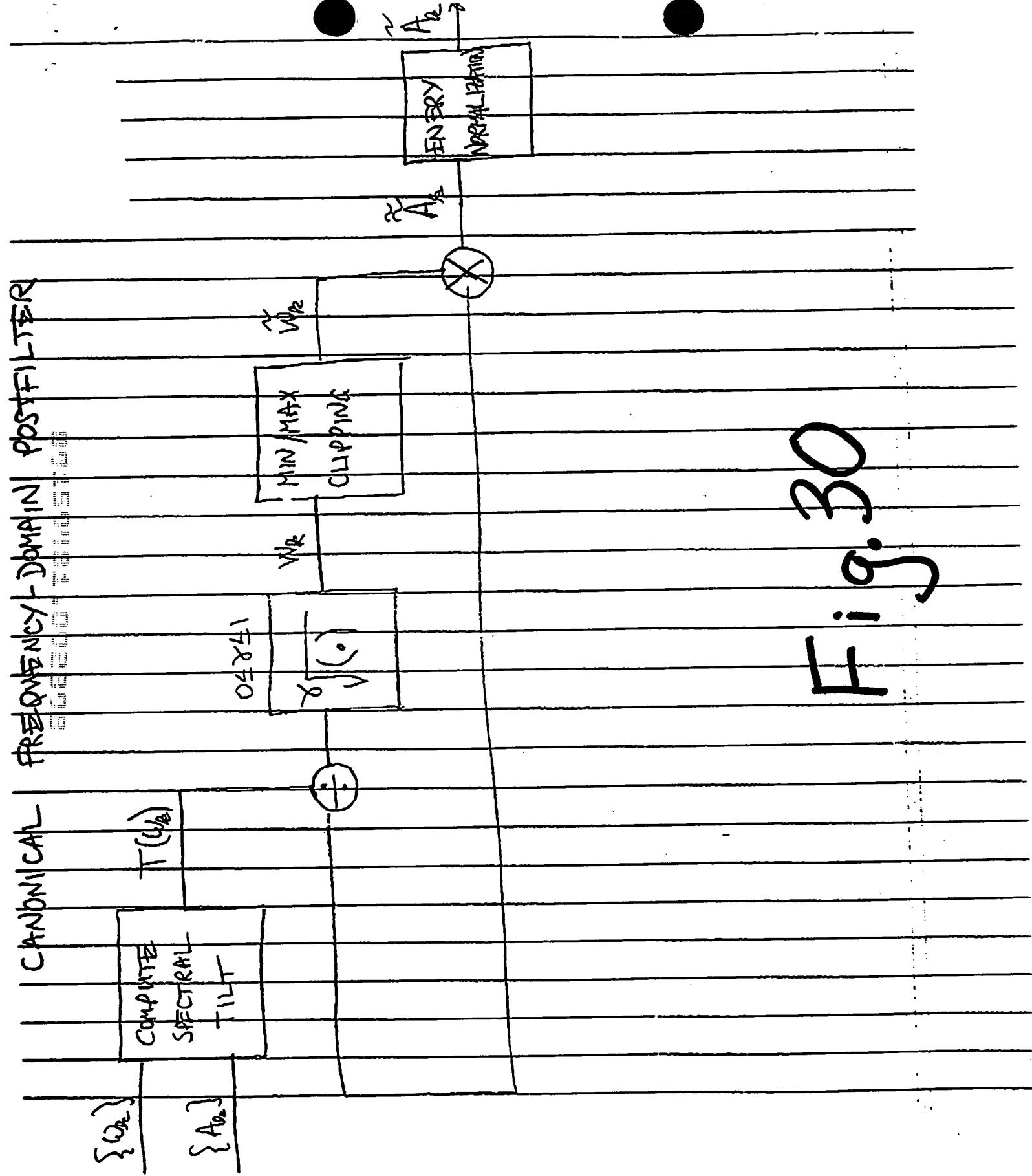
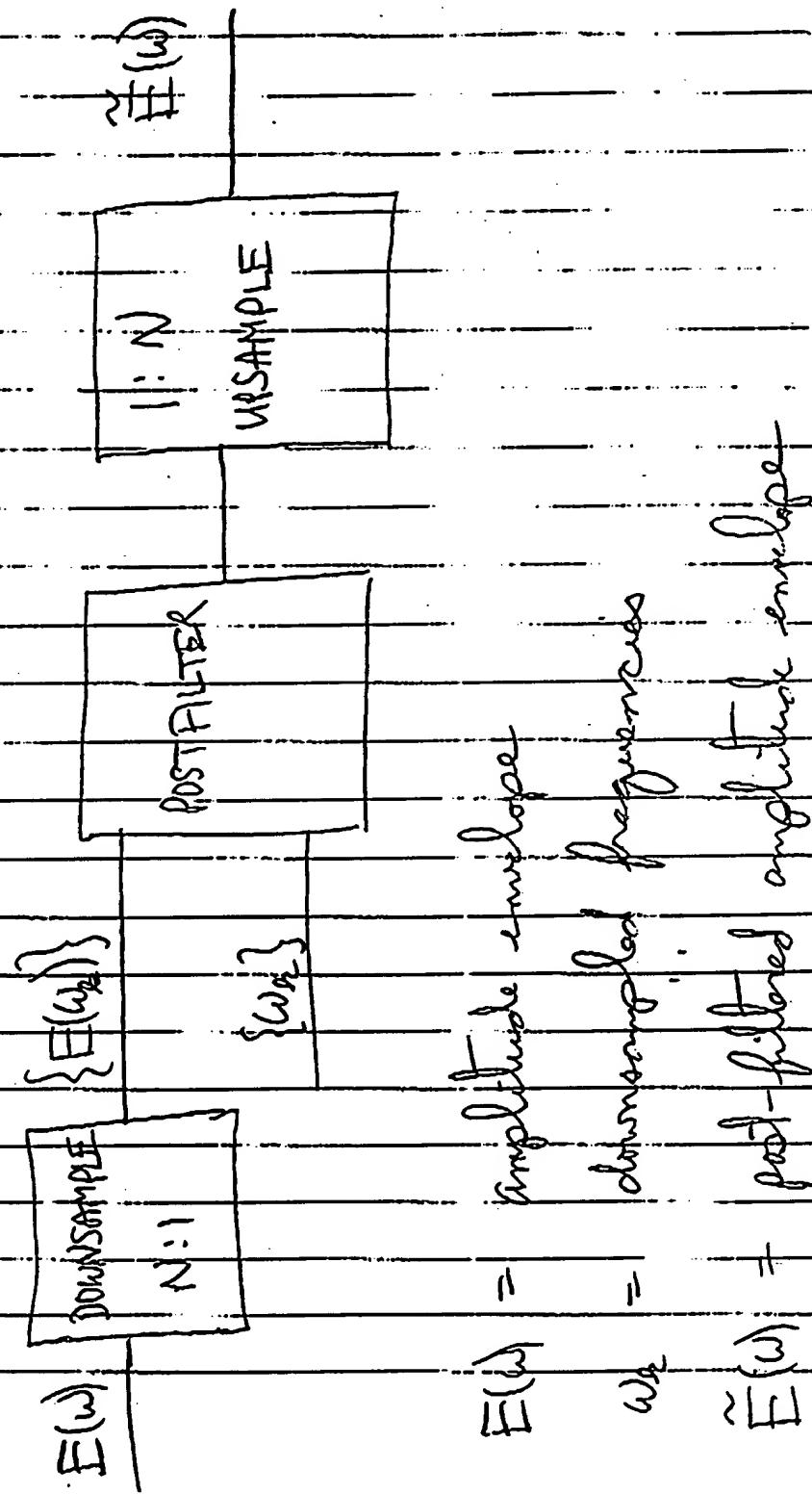


Fig. 30

constant component postfilter



$$\tilde{E}(w) = E(w) + \text{constant component}$$

= =

$$= E(w) + \text{constant component}$$

Fig 3

CONSTANT COMPLEXITY POST FILTER COMPUTED

FROM CEPSTRAL COEFFICIENTS

$$\{C_m\}$$

CEPSTRAL COEFFICIENTS

TO AMPLITUDE ENVELOPE

$$E(\omega)$$

CONSTANT COMPLEXITY
POST FILTER

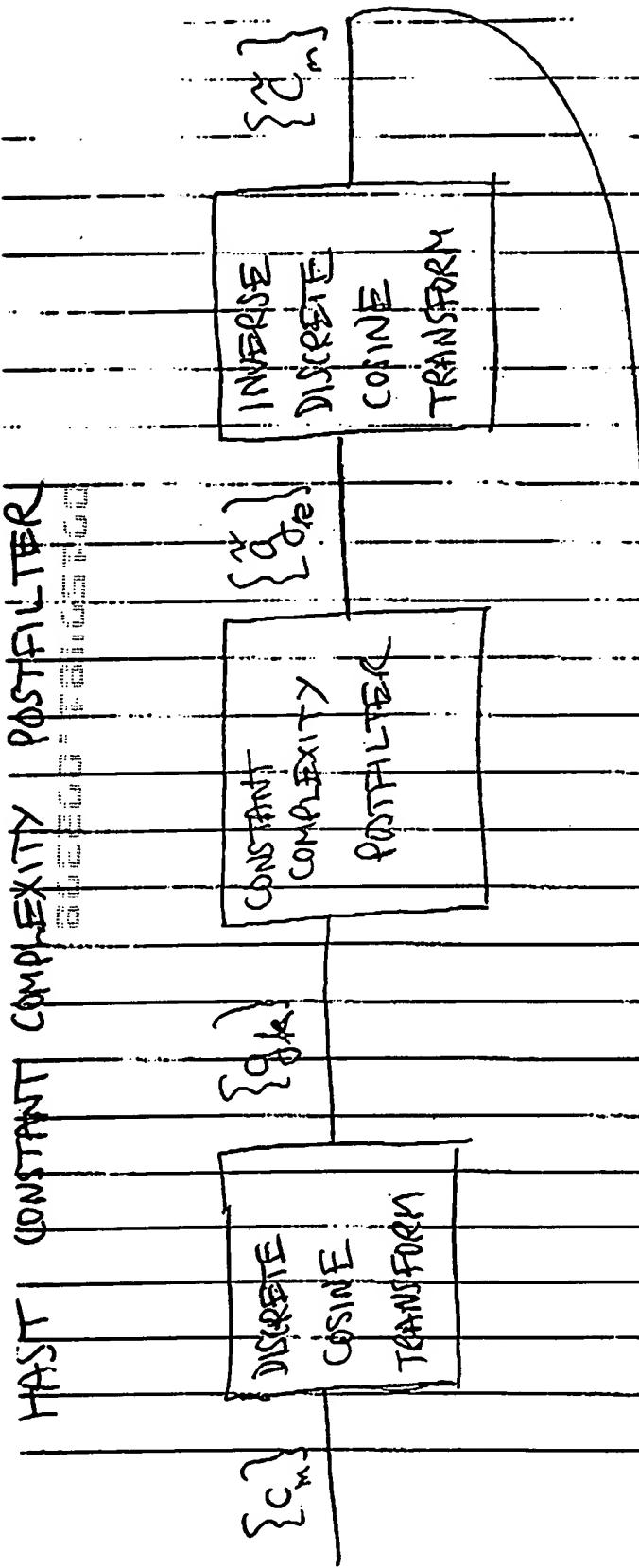
$$C_m = \sum_n e^{j\omega n} c_n$$

$$E(\omega) = \text{AMPLITUDE ENVELOPE}$$

$$E(\omega) = \text{AMPLITUDE ENVELOPE} + \text{POST-FILTERED}$$

$$\tilde{E}$$

Fig. 32



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COMPETITIVE ENVIRONMENT

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$$g_{\mu\nu} = \eta^{\mu\nu}$$

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لے ۳

efficient
capacitor

de + *green*

10

10

11

3

Chancery

23

— 8 —

100

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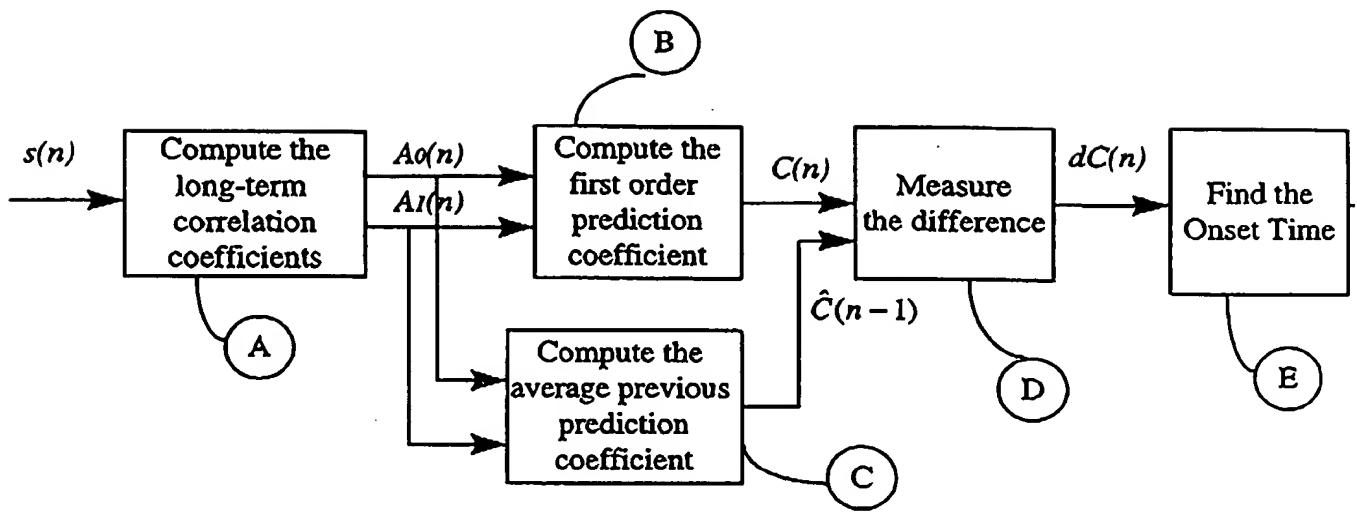


Figure 34.

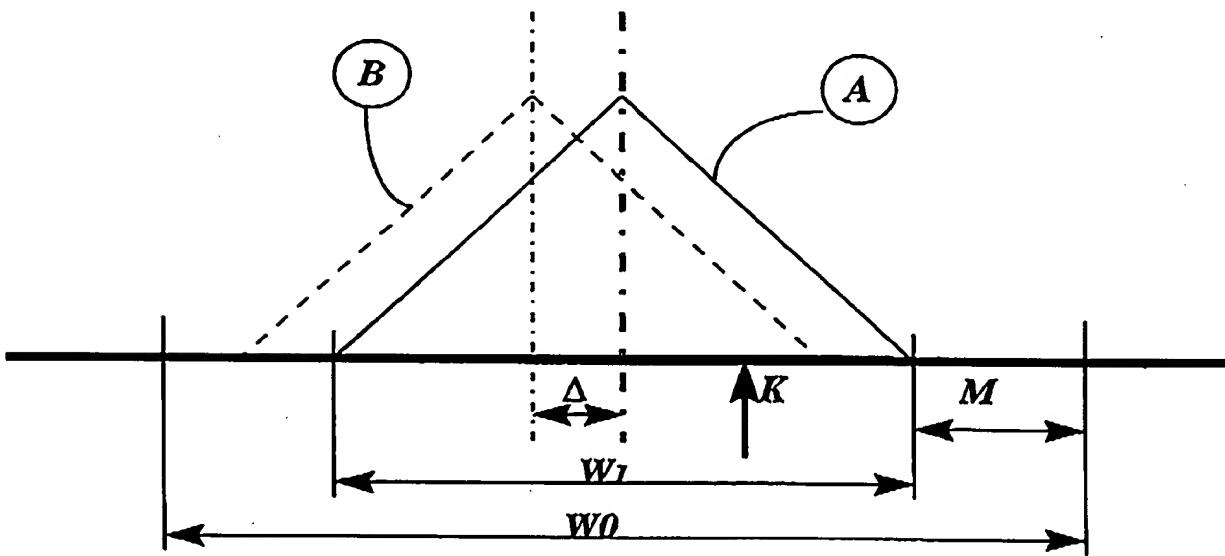


Figure 35.